BS EN 61883-6:2014

BSI Standards Publication

Consumer audio/video equipment — Digital interface

Part 6: Audio and music data transmission protocol

... making excellence a habit."

National foreword

This British Standard is the UK implementation of EN 61883-6:2014. It is identical to IEC 61883-6:2014. It supersedes BS EN [61883-6:2005](http://dx.doi.org/10.3403/30101599) which is withdrawn.

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Foreword

The text of document 100/2341/FDIS, future edition 3 of IEC [61883-6,](http://dx.doi.org/10.3403/02755535U) prepared by Technical Area 4 "Digital system interfaces and protocols" of IEC/TC 100 "Audio, video and multimedia systems and equipment" was submitted to the IEC-CENELEC parallel vote and approved by CENELEC as EN 61883-6:2014.

The following dates are fixed:

This document supersedes [EN 61883-6:2005](http://dx.doi.org/10.3403/30101599).

document have to be withdrawn

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In the official version, for Bibliography, the following notes have to be added for the standards indicated:

Annex ZA

(normative)

Normative references to international publications with their corresponding European publications

The following documents, in whole or in part, are normatively referenced in this document and are indispensable for its application. For dated references, only the edition cited applies. For undated references, the latest edition of the referenced document (including any amendments) applies.

NOTE 1 When an International Publication has been modified by common modifications, indicated by (mod), the relevant EN/HD applies.

NOTE 2 Up-to-date information on the latest versions of the European Standards listed in this annex is available here: [www.cenelec.eu](http://www.cenelec.eu/advsearch.html)

 $\frac{1}{2}$ $¹$ Superseded by IEEE 754:2008.</sup>

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CONSUMER AUDIO/VIDEO EQUIPMENT – DIGITAL INTERFACE –

Part 6: Audio and music data transmission protocol

1 Scope

This part of IEC 61883 describes a protocol for the transmission of audio and music data employing IEEE 1394 and specifies essential requirements for the application of the protocol.

This protocol can be applied to all modules or devices that have any kind of audio and/or music data processing, generation and conversion function blocks. This standard deals only with the transmission of audio and music data. The control, status and machine-readable description of these modules or devices should be defined outside of this standard according to each application area.

2 Normative references

The following documents, in whole or in part, are normatively referenced in this document and are indispensable for its application. For dated references, only the edition cited applies. For undated references, the latest edition of the referenced document (including any amendments) applies.

[IEC 60958](http://dx.doi.org/10.3403/00952384U) (all parts), *Digital audio interface*

IEC [60958-3](http://dx.doi.org/10.3403/02859696U), *Digital audio interface – Part 3: Consumer applications*

[IEC 61883-1](http://dx.doi.org/10.3403/01410651U), *Consumer audio/video equipment – Digital interface – Part 1: General*

IEEE 754:1985, *Standard for Binary Floating-Point Arithmetic*

IEEE 1394, *Standard for a High Performance Serial Bus*

IEEE 1394A, *Standard for a High Performance Serial Bus – Amendment 1*

3 Terms, definitions and abbreviations

3.1 Terms and definitions

For the purposes of this document, the terms and definitions given in IEC [61883-1](http://dx.doi.org/10.3403/01410651U), together with the following, apply.

3.1.1 32-bit floating-point data data type which is defined in IEEE 754

3.1.2 AM824 data 32-bit data consisting of an 8-bit label and 24-bit data

3.1.3

A/M protocol

protocol for the transmission of audio and music data over IEEE 1394

Note 1 to entry: IEEE 1394 describes the Audio and Music Data Transmission Protocol.

3.1.4

embedded synchronization clock

signal that carries information that is used by a sampling device to derive a sample clock

Note 1 to entry: In the context of A/M protocol, this synchronization clock is embedded in the SYT field of the CIP and carries timing information that refers to local CYCLE_TIME register values.

3.1.5

non-sampling device

devices that do not use clock timing in a way that may modify the analogue or digital audio signal

Note 1 to entry: Any clocks that the non-sampling devices use do not affect the accuracy of that data in normal operation. (See also [3.1.9,](#page-15-0) sampling device.)

3.1.6

reserved

keyword used to describe objects or the code values assigned to these objects

Note 1 to entry: Objects are: bit, byte, quadlet, octet, and field.

Note 2 to entry: The object or the code value are set aside for future standardization by the IEC.

3.1.7

sample clock

reference used at a sampling device to define the instant at which an audio data sample word is valid

Note 1 to entry: For over-sampled conversion systems, the sample clock is multiplied up to the over-sampling rate. Inside an asynchronous sampling frequency converter (ASFC), one sample clock is represented numerically by the relationship it has to another sample clock.

3.1.8

sample clock timing transfer

mechanism by which the sample clock of one device can be derived from a clock on another device such as by using an embedded synchronization clock

3.1.9

sampling device

device that depends on the timing of a sample clock to modify an audio signal in some way as it is being converted between the analogue and digital domains, or between two independent sampling frequencies

Note 1 to entry: Examples of a sampling device are an analogue-to-digital converter (ADC), a digital-to-analogue converter (DAC) and an ASFC.

3.1.10 sampling frequency $F_{\rm s}$ frequency of the sample clock

3.1.11 stream uni-directional data transmission

3.1.12

synchronization clock frequency

*F*sync

embedded synchronization clock frequency using the A/M protocol has to be less than the isochronous cycle rate of 8 kHz

Note 1 to entry: The rate is defined as follows:

 F_{sync} = F_{s} / SYT_INTERVAL

The SYT_INTERVAL value is defined in the CIP header for each sampling frequency.

3.1.13

synchronization clock source

device that supplies an embedded synchronization clock that another device uses to derive a sample clock

Note 1 to entry: A synchronization clock source does not need to be a source device for audio data.

3.1.14 synchronization clock destination

3.1.15.1 clock jitter definition

deviation in the timing of clock transitions when compared to an ideal clock

Note 1 to entry: The ideal clock can be considered to have a frequency of exactly the same long-term average frequency and aligned for zero mean phase offset from the real clock. For a sample clock, the jitter amplitude defined in this way is directly related to the amplitude of the jitter modulation products produced in a sampling device.

3.1.15.2

embedded synchronization clock jitter

jitter in the embedded synchronization clock includes the effect of errors (including limited precision) in the embedded SYT data and jitter in the CYCLE_TIME register used to decode the SYT

3.1.15

time stamp

quantized timing in which an event occurs based on a reference clock

Note 1 to entry: The reference clock is CYCLE_TIME unless otherwise specified in this standard.

3.2 Abbreviations

4 Reference model for data transmission

4.1 General

This clause describes a reference model for data transmission.

Figure 1 – Reference model for audio and music data transmission

Figure 2 – Reference model for AM824 data transmission

[Figure 1](#page-17-2) gives an outline for audio data transmission from a transmitter to a receiver. It has four major layers denoted as CIP (common isochronous packet) layer, packetization layer, adaptation layer and application layer.

4.2 Application layer

Each application defines its own application sequence and the interface to the adaptation layer. The application sequence in [Figure 1](#page-17-2) is data in a format such as an audio signal format. The Nominal_Sampling_Frequency is the ideal sampling frequency for the application sequence. The range of *Sampling_Frequency* should be defined by the application. The audio signal at Nominal_Sampling_Frequency can be reproduced at the actual rate of Sampling Frequency in operation. This means that the value of Sampling Frequency may have some deviation and/or may vary in time in contrast with Nominal Sampling Frequency.

Additional information in [Figure 1](#page-17-2) is any information other than events of a sequence (audio samples) being transmitted at a given rate.

4.3 Adaptation layer

The adaptation layer defines a process to convert an application sequence to an event sequence and vice versa. The conversion process may not be required if an application sequence and an event sequence have the same structure. If an event sequence consists of events of 24-bit payload, such as AM824 data defined in [8.2,](#page-28-0) and if the bit length of an audio sample of the application sequence is not 24-bit, some conversion between Sampling_Frequency and Transfer_Frequency may be required (see [Figure 2](#page-18-2) and Clause [11\)](#page-43-1). The Transfer Frequency represents the frequency of occurrence of a data block, which is equivalent to a cluster event. The Transfer_Frequency is used for describing a conceptual transmission model.

The transfer rate of an event sequence is 24 \times Transfer Frequency [bit/s] in the case of AM824.

Generally, the adaptation layer is designed in such a way that both the application sequence at Sampling_Frequency and its Nominal_Sampling_Frequency are carried. In this specification, Nominal_Sampling_Frequency, which would usually be one of the ancillary data items, is carried by SFC (sampling frequency code) which is defined in Clause [10.](#page-40-0) The information in Nominal Sampling Frequency is necessary for using command based rate control or making a copy. On the other hand, Sampling Frequency is necessary for clockbased rate control. Although Sampling Frequency is not explicitly transmitted, it can be estimated from SYT_INTERVAL and time stamps by the algorithm specified for the AM824 data type.

An application specification defines the process (shown in the grey shaded area of [Figure 1\)](#page-17-2) to convert the application's signal (application sequence) to an event sequence. This standard assumes that the application specification is an external document using the definition of an event sequence for the adaptation process. For several generic data types this standard also defines the adaptation layer.

The adaptation to an event sequence is the point at which the packetization process interfaces to the application. The packetization process can be described as IEEE 1394 adaptation from the point of view that the data stream utilizes IEEE 1394 as its transport.

More details of this layer are described in Clause [12.](#page-48-0)

4.4 Packetization layer

The AM824 sequence is directly packetized to CIP or depacketized from CIP in the packetization layer.

The Transfer Frequency can be implicitly expressed by the output of a locked PLL circuit, as shown in [Figure 3,](#page-19-1) instead of being explicitly denoted in the packetization layer.

Figure 3 – Implementation example of receiver

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5 Transport requirements

5.1 Arbitrated short bus reset

All modules or devices which implement this A/M protocol should have the capability of "arbitrated short-bus reset" in order to prevent the interruption of audio and music data transmission when a bus reset occurs.

5.2 Bit, byte, and quadlet ordering

This standard adopts the ordering of bit, byte, and quadlet for bus packets in accordance with IEEE 1394.

6 Packet header for audio and music data

6.1 General

This clause defines the packet format in the CIP layer described in [Figure 1.](#page-17-2)

6.2 Isochronous packet header format

The header for an isochronous packet which conforms to the A/M protocol shall have the same format given in [Figure 4,](#page-20-7) which is part of the isochronous packet format defined in IEEE 1394.

Figure 4 – Isochronous packet header

The isochronous packet header fields are defined with unique values that are specified in [Table 1.](#page-20-8)

6.3 CIP header format

[IEC 61883-1](http://dx.doi.org/10.3403/01410651U) defines a two-quadlet CIP header for a fixed length source packet with SYT field, repeated here for clarity as in [Figure 5.](#page-21-3) The CIP header format for an isochronous packet, which conforms to the audio and music data transmission protocol, shall use the CIP header.

Figure 5 – Common isochronous packet (CIP) format

[Table 2](#page-21-4) defines the fields with unique values that are specified by this protocol.

Table 2 – CIP fields

7 Packetization

7.1 Packet transmission method

When a non-empty CIP is ready to be transmitted, the transmitter shall transmit it within the most recent isochronous cycle initiated by a cycle start packet. The behaviour of packet transmission depends on the definition of the condition in which "a non-empty CIP is ready to be transmitted". There are two situations in which this condition is defined.

- a) In order to minimize TRANSFER DELAY, the condition of a non-empty CIP being ready for transmission is defined to be true if one or more data blocks have arrived within an isochronous cycle. This transmission method is called non-blocking transmission and is described in detail in [7.4.](#page-22-1)
- b) The condition of "non-empty CIP ready" can also be defined as true when a fixed number of data blocks have arrived. This transmission method is called blocking transmission and is described in [Annex A.](#page-109-0)

7.2 Transmission of timing information

A CIP without a source packet header (SPH) has only one time stamp in the SYT field. If a CIP contains multiple data blocks, it is necessary to specify which data block of the CIP corresponds to the time stamp.

The transmitter prepares the time stamp for the data block which meets this condition:

$$
mod(data block count, SYT_INTERVAL) = 0
$$
 (1)

where

data block count is the running count of transmitted data blocks;

SYT_INTERVAL denotes the number of data blocks between two successive valid SYTs, which includes one of the data blocks with a valid SYT. For example, if IEC 61883-6:2014 © IEC 2014 – 21 – BS EN 61883-6:2014

there are three data blocks between two valid SYTs, then the SYT_INTERVAL would be 4.

The receiver can derive the index value from the DBC field of a CIP with a valid SYT using the following formula:

index = **mod**((SYT_INTERVAL − **mod**(DBC, SYT_INTERVAL)), SYT_INTERVAL (2)

where

index is the sequence number;

SYT_INTERVAL denotes the number of data blocks between two successive valid SYTs, which includes one of the data blocks with a valid SYT;

DBC is the data block count field of a CIP.

The receiver is responsible for estimating the timing of data blocks between valid time stamps. The method of timing estimation is implementation-dependent.

7.3 Time stamp processing

A data block contains all data arriving at the transmitter within an audio sample period. The data block contains all data which make up an "event".

The transmitter shall specify the presentation time of the event at the receiver. A receiver for professional use shall have the capability of presenting events at the time specified by the transmitter. A consumer-use or cost-sensitive receiver is not required to support this presentation-time adjustment capability.

If a function block receives a CIP, processes it and subsequently re-transmits it, the SYT of the outgoing CIP shall be the sum of the incoming SYT and the processing delay.

The transmitter shall add TRANSFER_DELAY to the quantized timing of an event to construct the SYT. The TRANSER_DELAY value is initialized with the DEFAULT_TRANSFER_DELAY value. For professional use, TRANSFER DELAY may be changed to achieve a shorter TRANSFER_DELAY value, according to the bus configuration. Products for consumer use are not required to support the modification of TRANSFER_DELAY.

The DEFAULT TRANSFER DELAY value is $(354,17 + 125)\mu s$, which accommodates the maximum latency time of CIP transmission through an arbitrated short bus reset.

7.4 Transmission control

7.4.1 Non-blocking transmission method

[Figure 6](#page-23-0) illustrates the non-blocking transmission method.

Figure 6 – Non-blocking transmission method

The transmitter shall construct a packet in every nominal isochronous cycle. Each packet shall comply with the following constraint:

$$
0 \leq N \leq SYT_N \text{INTERVAL} \tag{3}
$$

where *N* is the number of events in the packet.

In normal operation the transmitter shall not transmit events late and shall not transmit packets early. The resulting conditions may be expressed as follows:

Packet_arrival_time_L <= Event_arrival_time[0] + TRANSFER_DELAY (4)

$$
Event_arrival_time[N-1] \leq Packet_arrival_time_F
$$
 (5)

where

[Figure 7](#page-24-1) illustrates the transmission control rules as described in Clause [7.](#page-21-0)

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Figure 7 – Transmission parameters

In the event of lost opportunities to transmit non-blocking packets, a method of catching up may be provided (refer to [Annex C\)](#page-112-0).

7.4.2 Blocking transmission method

[Figure 8](#page-24-2) shows a diagram of a blocking transmission method.

The blocking method may be used by a transmitter, which has only the ability to transmit packets of the same size. In order to indicate "no data", the transmitter may transmit an empty packet or a special non-empty packet which has the "NO-DATA" code in its FDF and has the same size of dummy data as a non-empty packet. The transmitter shall set the time stamp of the first data block in a packet.

For blocking, the duration of the successive events in a CIP have to be added to DEFAULT_TRANSFER_DELAY.

If a CIP contains *N* audio samples of a stream at Sampling Transmission Frequency (STF), then:

TRANSFER_DELAY >= DEFAULT_TRANSFER_DELAY + 1/STF × *N* × 1 000

where

TRANSFER DELAY is the latency of transmission;

DEFAULT_TRANSFER_DELAY is the initialized value of TRANSFER_DELAY;

STF is the sampling transmission frequency:

N is the number of audio samples in a CIP.

The TRANSFER DELAY for each STF when DEFAULT TRANSFER DELAY = 479,17 us $(= 354.17 \text{ }\mu\text{s} + 125 \text{ }\mu\text{s})$ is defined in [Table](#page-39-1) 21.

It is recommended that the receiver have 250 µs of extra buffer.

8 Event types

8.1 General

All the subformats described in this standard shall use only 32-bit aligned events.

If multiple event sequences are synchronized, it is possible to convert the sequences into a single event which consists of an ordered collection of those sequences which occurred at the same time. The ordered collection is called a cluster. A cluster consists of ordered units. In the case of data*,* a unit consists of a single sequence. In the case of a pack, the unit may consist of several sequences packed together. The number of units in a single cluster is called the dimension, and is denoted by CLUSTER_DIMENSION. [Figure 9](#page-26-0) illustrates these concepts.

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Figure 9 – Cluster events

In order to efficiently cluster non-32-bit aligned sequences which occur at the same time, the pack event type is defined. For example, four events of 24-bit data can be collected into a pack of three quadlets.

An event which is neither a cluster nor a pack is simply called data.

Only the pack and data types can be combined into units to make a cluster. All events in a cluster shall be of the same type.

UNIT SIZE is the number of quadlets in a unit.

UNIT_DIMENSION is the number of sequences in a unit.

The UNIT_DIMENSION of data is always 1.

[Figure](#page-27-0) 10 illustrates pack and cluster events.

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Figure 10 – Pack and cluster events

[Figure](#page-27-1) 11 illustrates the structure of a pack which consists of four 24-bit event sequences $(UNIT_DIMENSION = 4, UNIT_SIZE = 3).$

Figure 11 – Pack event with 24-bit event sequence

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$$
-27-
$$

Since the cluster is an abstract event, only pack or data shall be specified as an event type for a subformat. However, the DBS shall reflect the size in quadlets of all cluster events in a data block. In the case of a clustered sequence:

$$
DBS = \sum_{n=0}^{(\text{clusters}-1)} (Unit_Size_n \times \text{CLUSTER_DIMENSION}_n)
$$
 (6)

where

Generally, the number of elementary sequences in a CIP is given by the following:

number of sequences =
$$
DBS \times UNIT_DIMENSION / UNIT_SIZE
$$
 (7)

For the pack illustrated in [Figure 9](#page-26-0) and [Figure](#page-27-0) 10, DBS = 6, CLUSTER DIMENSION = 2, UNIT DIMENSION = 4, UNIT SIZE = 3.

The number of successive events in a CIP is equal to the number of data blocks in a CIP and given by

$$
NEVENTS_SUCCESSIVE = (data_length / 4 - CIPH_SIZE) / DBS
$$
 (8)

where

The ordering of sequences in an event is application-specific and is not within the scope of this standard. For example, the identification of audio channels in a multichannel transmission will be defined elsewhere.

8.2 AM824 data

8.2.1 Generic format

[Figure](#page-28-2) 12 shows the AM824 format that is used in the following conditions and [Table 3](#page-29-2) provides a LABEL definition.

NOTE A 32-bit data consisting of an 8-bit label and 24-bit data are called AM824 data.

UNIT $SIZE = 1$ quadlet/unit UNIT_DIMENSION = 1 sequence/unit

Figure 12 – Generic AM824 format

A receiver capable of processing AM824 data shall check the label for each AM824 data in a sequence being received.

Figure 13 – AM824 data with SUB LABEL

If an application requires many data types, SUB LABEL, as shown in [Figure](#page-29-0) 13, may be used to extend the number of data types defined by LABEL, see [Table 3.](#page-29-2) An application map representation for [Table 3](#page-29-2) is shown in [Figure](#page-29-1) 14.

Value	Description	
$00_{16} - 3F_{16}$	IEC 60958 conformant	
$40_{16} - 4F_{16}$	Multi-bit linear audio	
$50_{16} - 57_{16}$	One-bit audio (plain)	
$58_{16} - 5F_{16}$	One-bit audio (encoded)	
$60_{16} - 67_{16}$	High-precision multi-bit linear audio	
$70_{16} - 7F_{16}$	Reserved	
$80_{16} - 83_{16}$	MIDI conformant	
$84_{16} - 87_{16}$	Reserved	
$88_{16} - 8B_{16}$	SMPTE time code conformant	
$8C_{16} - 8F_{16}$	Sample count	
$90_{16} - BF_{16}$	Reserved	
$CO_{16} - EF_{16}$	Ancillary data	
$F0_{16} - FF1_{6}$	Reserved	

Table 3 – LABEL definition

Figure 14 – AM824 LABEL allocation map

8.2.2 [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data

IEC [60958](http://dx.doi.org/10.3403/00952384U) confomant data format is shown in [Figure](#page-30-2) 15.

Figure 15 – [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data format

All information defined in [IEC 60958](http://dx.doi.org/10.3403/00952384U) is mapped into the data format shown in [Figure](#page-29-1) 14 and [Table 4.](#page-30-4) For each [IEC 60958](http://dx.doi.org/10.3403/00952384U) frame, both sub-frames shall be transmitted together in the same event. The corresponding quadlet may be consecutive or non-consecutive. If multiple [IEC 60958](http://dx.doi.org/10.3403/00952384U) streams are transmitted, then their sub-frames shall not be interleaved. Applications which use this data type shall follow [IEC 60958](http://dx.doi.org/10.3403/00952384U).

8.2.3 Multi-bit linear audio (MBLA)

Multi-bit linear audio (MBLA) is shown in [Figure](#page-30-3) 16.

Figure 16 – MBLA data

The label field of MBLA has two fields for ASI (application-specific information). The definition of ASI2 depends on the ASI1 value described in [Table 5.](#page-30-5)

Table 5 – ASI1 definition

[Figure](#page-31-1) 17 shows raw audio format and [Table 6](#page-31-3) provides a VAL definition.

Figure 17 – Raw audio data

Table 6 – VBL (valid bit length code) definition

Value	Description
00 ₂	24 bit
01 ₂	20 bit
10 ₂	16 bit
11 ₂	Reserved

The audio data shall be expressed in 24-bit 2's complement format. If the data active word length is less than 24 bit, the correct number of zero bit shall be padded below the least significant bit to make a 24-bit data structure.

For example, a 20-bit audio data shall be placed in a 24-bit field as shown in [Figure](#page-31-2) 18 (note the four zero pad bit at the right end of the structure).

Figure 18 – Alignment of 20-bit data in 24-bit field

For audio data word lengths of less than 24 bit the VBL indication can be used by receivers to determine if the data can be truncated to less than 24 bit without changing the value. If the word length is not known or variable the data should be aligned at the most significant bit and the VBL code for 24-bit indication should be used.

8.2.4 One-bit audio

One-bit audio defines its own sampling frequency code (SFC). [Table 7](#page-32-3) and [Table 8](#page-32-4) provide LABEL definitions for one-bit audio (plain) and one-bit audio (encoded), respectively.

Table 7 – LABEL definition for one-bit audio (plain)

Table 8 – LABEL definition for one-bit audio (encoded)

8.2.5 MIDI conformant data

[Figure](#page-32-1) 19 illustrates a MIDI conformant data format and [Table 9](#page-32-5) provides a C definition.

Figure 19 – MIDI conformant data format

Table 9 – C (counter) definition

If the CIP carries only MIDI conformant data or cluster, and there are no MIDI data to be packed into a CIP, the packet should be an empty packet rather than a packet of all no-data codes.

The no-data code defined in MIDI conformant data may be used as no data for other AM824 data types if necessary. The usage of no data described above should be applied to the AM824 data types which use no data.

[Figure](#page-32-2) 20 illustrates the no-data structure.

Figure 20 – No-data format

Successful implementation of MIDI conformant data may require additional information. Attention is drawn to MMA/AMEI Recommended Practice 027.

8.2.6 SMPTE time code data

The SMPTE time code is defined in 1394 Trade Association document 1999024, SMPTE Time Code and Sample Count Transmission Protocol Version 1.0.

8.2.7 Sample count data

The Sample count transmission is defined in 1394 Trade Association document 1999024, SMPTE Time Code and Sample Count Transmission Protocol Version 1.0.

8.2.8 High-precision multi-bit linear audio

The multi-bit linear audio (MBLA) is limited to sample words up to 24 bit length. Linear PCM audio data longer than 25 bit length and up to 196 bit length can be transmitted with highprecision multi-bit linear audio, see [Figure](#page-33-3) 21.

Figure 21 – High-precision multi-bit linear audio data

The high-precision multi-bit linear audio data use the LABEL from 60_{16} to 67_{16} . The label field of the high-precision multi-bit linear audio has Num. (slot number) field. The definition of Num. field is given in [Table](#page-33-4) 10.

Value	Description
000 ₂	1st slot number (Num. $= 0$)
001 ₂	2nd slot number (Num. $= 1$)
010 ₂	3rd slot number (Num. $= 2$)
\cdots	\cdots
111_2	8th slot number (Num. = 7)

Table 10 – Num. (slot number) definition

The high-precision multi-bit linear audio data longer than 25 are divided into more than 2 quadlet sequence slots. The Num. (slot number) shall start with Num. = 0 (LABEL = 60_{16}) and be sequential. [Figure](#page-34-1) 22 shows generic quadlet sequence for the high-precision multi-bit linear audio data.

Figure 22 – Generic high-precision quadlet sequence

8.2.9 Ancillary data

8.2.9.1 Generic ancillary data

Generic ancillary data is illustrated in [Figure](#page-34-2) 23 and the definition of LABEL is probided in [Table](#page-34-3) 11. The definition of Byte 1, Byte 2, Byte 3 and transmission method, timing accuracy and interval, for instance, should be given by each instance of ancillary data. It is recommended that all information carried by ancillary data should be transmitted repeatedly in a reasonably short interval of time while the information is valid so that the receiver does not have to wait for the information. It is recommended that Byte 1 is defined as a SUB LABEL that specifies Byte 2 and Byte 3.

Figure 23 – Generic ancillary data

Table 11 – LABEL definition for ancillary data type

8.2.9.2 Common ancillary data

8.2.9.2.1 General

Common ancillary data carries information common to all applications under a category of such as copyright information. The usage of these data are described in [11.4.2.3.](#page-46-1) [Table](#page-34-4) 12 provides LABEL definition for common ancillary data.

Table 12 – LABEL definition for common ancillary data

8.2.9.2.2 Ancillary no data

Ancillary no-data, shown in [Figure](#page-35-0) 24, provides a no-data event only for AM824 data that does not define its own no data. [Table](#page-35-1) 13 provides a CONTEXT definition. AM824 data types that define their own no data shall not use this ancillary no data.

In order to determine whether the AM824 data type carries valid information, it is required that no data specifies the AM824 data type to which it belongs. For this reason, the AM824 data type derived from a given no data should be identical to the AM824 data that carries valid information. Subclause [8.2.5](#page-32-0) allows the use of no data defined in MIDI conformant data.

Figure 24 – Ancillary no data

Table 13 – CONTEXT definition

8.2.9.2.3 ASID

ASID (Audio Software Information Delivery) defines transmission methods of ISRC, UPC/EAN and content usage (copyright assertion) information carried by the AM824 data.

The general format for ASID is shown in [Figure](#page-36-2) 25.
LABEL $=$ C.O ∪∪16	ABEL . .	` data

Figure 25 – General format for ASID

The second byte SUB LABEL following the LABEL identifies the particular type of ASID data as shown in [Table](#page-36-0) 14.

For details, see the ASID specification.

8.2.10 Application specific ancillary data

Application specific ancillary data carries information specific to an application, which is transmitted along with the audio and music data. Examples are mapping of sequence of a compound data block to speaker location, microphone location, or signal name.

The general format for the application-specific ancillary data is shown in [Figure](#page-36-1) 26.

Figure 26 – General format for application-specific ancillary data

The first byte ("LABEL") indicates that these data are for the application specific ancillary data of the type shown in [Table](#page-36-2) 15. The second byte ("SUB LABEL") further identifies the particular data that follows. For details, refer to [12.2](#page-63-0) for DVD-Audio, [12.3](#page-67-0) for SACD, [12.4](#page-71-0) for Blu-ray and [12.5](#page-80-0) for Multi-bit Linear Audio (MBLA).

8.3 32-bit floating-point data

This data type carries 32-bit floating-point data defined in IEEE 754-1985, Standard for Binary Floating-Point Arithmetic.

UNIT $SIZE = 1$ quadlet/unit UNIT_DIMENSION = 1 sequence/unit

[Figure](#page-37-0) 27 illustrates the structure of 32-bit floating-point data.

8.4 24-bit × **4 audio pack**

UNIT $SIZE = 3$ quadlet/unit UNIT_DIMENSION = 4 sequences/unit

[Figure](#page-37-1) 28 illustrates the structure of a 24-bit \times 4 audio pack.

Figure 28 – 24-bit × **4 audio pack format**

W1, W2, W3, W4: 24-bit raw audio data

The audio data shall be expressed in 24-bit 2's complement. In case of less than 24 bit, the correct number of zero bit shall be padded below the least significant bit to make a 24-bit data structure. For an example of this, refer to [8.2.3.](#page-30-0)

8.5 32-bit generic data

UNIT_SIZE = 1 quadlet/unit UNIT_DIMENSION = 1 sequence/unit

[Figure](#page-37-2) 29 illustrates the structure of 32-bit generic data.

Figure 29 – 32-bit generic data format

9 FDF definition

9.1 Overview

Under the A/M packet format as described in [6.3,](#page-20-0) format dependent field (FDF) is used for specifying subformat type and additional information described in Clause [4.](#page-17-0) [Table](#page-37-3) 16 defines the subformat and FDF allocations.

Table 16 – Subformat and FDF allocations

Each subformat may use a cluster for synchronized multiple sequences unless otherwise specified.

9.2 Basic format

[Table](#page-38-0) 17 and [Table](#page-38-1) 18 show the DBS for AM824 and 32-bit floating-point data and DBS for 24-bit \times 4 audio pack, respectively.

Table 17 – DBS for AM824 and 32-bit floating-point data

Table 18 – DBS for 24-bit \times 4 audio pack

[Figure](#page-38-2) 30 illustrates a generic FDF definition. [Table](#page-38-3) 19 and [Table](#page-39-0) 20 provide EVT and SFC definitions, respectively. The TRANSFER_DELAY for the given SFC definition is provided in [Table](#page-39-1) 21.

Figure 30 – Generic FDF definition

Table 19 – Event type (EVT) code definition

Table 20 – Default SFC table

If a packet of AM824 data contains only [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data and a transmitter functions as a gateway, the transmitter should estimate the sample transmission frequency for the SFC rather than copying the sampling frequency code embedded in the original [IEC 60958](http://dx.doi.org/10.3403/00952384U) data.

Equation (9) can be used to determine the required bus bandwidth allocation. The required isochronous bandwidth is given below:

BW = (int (max(*FS*) / 8000) + 1) ×
$$
\sum_{n=0}^{\text{clusters-1}} (UNIT_SIZE_n \times CLUSTER_DIMENSION_n) \times 8000
$$
 (9)

where

9.3 Special format

Figure 31 – FDF code for NO-DATA packet

The transmitter shall use the FDF code shown in [Figure](#page-40-0) 31 when a packet is a no-data packet only for blocking transmission. The transmitter shall not use this FDF code for non-blocking transmission. The receiver shall ignore all the data in a CIP with this FDF code.

10 FDF definition for AM824 data

10.1 Definition of N-flag

Figure 32 – Structure of FDF for AM824 data type

The N-flag as shown in [Figure](#page-40-1) 32 shall be used to select the AM824 LABEL space and the adaptation process described in [10.4.](#page-42-0)

Any AM824 data type shall occupy the same space in both LABEL spaces. An application may use only one of two LABEL spaces by giving a fixed value to the N-flag. Only an AM824 data type that owns the LABEL space or application-specific ancillary data, which is defined in [8.2.10,](#page-36-3) can inhibit the use of one of the LABEL spaces.

10.2 Supplementary SFC definition

In [IEC 61883-6:2002](http://dx.doi.org/10.3403/02755535), there is only one SFC table that specifies both Nominal Sampling Frequency and SYT_INTERVAL.

For clarification, and in order to enable backwards compatibility, in this standard, the SFC definition is changed so that a new AM824 data type, which was defined in [IEC 61883-6:2002](http://dx.doi.org/10.3403/02755535), may define its own SFC table. In order to maintain compatibility with [IEC 61883-6:2002,](http://dx.doi.org/10.3403/02755535) in the case of FDF = 0000 0xxx₂, the default SFC table shall be identical to the table defined in [IEC 61883-6:2002](http://dx.doi.org/10.3403/02755535). Only a new AM824 data type may override the default SFC table. [Figure](#page-41-0) 33 shows an SFC interpretation and [Figure](#page-41-1) 34 gives an example of an FDF for AM824 and AM824 LABEL space.

The empty packet defined in [IEC 61883-1](http://dx.doi.org/10.3403/01410651U) shall use the default SFC table.

Figure 34 – FDF for AM824 and AM824 LABEL space

10.3 Clock-based rate control mode (FDF = 0000 0xxx₂)

10.3.1 Introductory remark

This FDF value, defined in [IEC 61883-6:2002,](http://dx.doi.org/10.3403/02755535) is interpreted to indicate that the data transmission rate is controlled by a transmission clock reproduced by means of a timestamp.

The meaning of this FDF value, as specified in [10.3.2,](#page-41-2) is not changed.

10.3.2 Default SFC table for (FDF = 0000 0xxx₂)

When FDF has 0000 $0xxx₂$, default SFC is defined in [Table](#page-42-1) 22.

Table 22 – Default SFC table for FDF = 0000 0xxx₂

The TRANSFER_DELAY for blocking transmission, in the case of DEFAULT_TRANSFER_ DELAY = 479,17 μ s = (354,17 + 125) μ s, corresponds to the default SFC table as given in [Table](#page-42-2) 23.

Table 23 – TRANSFER_DELAY for blocking transmission

Value	TRANSFER DELAY
	μs
00_{10}	$479.17 + 250.00 = 729.17$
01_{10}	$479.17 + 181.41 = 660.58$
02_{10}	$479.17 + 166.67 = 645.84$
03_{10}	$479.17 + 181.41 = 660.58$
04_{10}	$479.17 + 166.67 = 645.84$
05_{10}	$479.17 + 181.41 = 660.58$
06_{10}	$479.17 + 166.67 = 645.84$
07_{10}	Reserved

10.4 Command-based rate control mode (FDF = 00001xxx₂)

10.4.1 Introductory remark

This FDF value indicates that the data transmission rate is controlled by a command set such as the AV/C command set for the rate control of the isochronous data flow.

This transmission mode can be used for reproducing an application sequence at a receiver or for high-speed data transfer without using a timestamp in the SYT field.

If the timing information is available, the transmitter should provide the correct timestamp in the SYT field according to the integer multiplier *n* so that the clock-based rate controlled receiver can receive the data transmitted in this mode.

SYT_INTERVAL_{N-flag=1} = SYT_INTERVAL_{N-flag=0} \times *n* (*n* >= 1)

where SYT_INTERVAL_{N-flag=1} and SYT_INTERVAL_{N-flag=0} denote SYT_INTERVAL specified by the SFC table in the cases in which FDF = 0000 1xxx $_2^{\circ}$ and FDF = 0000 0xxx $_2$, respectively. The integer multiplier *n* is obtained by a command.

10.4.2 Default SFC table for (FDF = 0000 1xxx₂)

When FDF has 0000 1xxx $_2$, default SFC is defined in [Table](#page-43-0) 24.

Table 24 – Default SFC table for FDF = 0000 1xxx₂

The DBS of an event is independent of the transfer speed.

11 AM824 adaptation process

11.1 Introductory remark

This clause describes typical methods of adaptation to an AM824 sequence.

11.2 Basic sequence conversion

Transfer Frequency is identical to Sampling Frequency (transfer frequency of the application sequence such as audio) to be packetized if each event in the application sequence (each audio sample) is stored in one unit such as one AM824 data of an AM824 sequence.

Figure 35 – Adaptation to AM824 sequence

[Figure](#page-43-1) 35 describes an example of an adaptation process in which each event of the application sequence is 8 bit in length and three events are stored in a single AM824 data which has a 24-bit payload. In this case, the relation between Sampling Frequency and Transfer_Frequency is expressed by

Sampling Frequency = $L \times$ Transfer Frequency

where $L = 3$.

The parameters Sampling_Frequency, Transfer_Frequency and *L* cannot be specified independently. All of them are specified by the SFC code selected by the AM824 data type.

11.3 Sequence multiplexing

If the event occurrence rate of an application sequence is less than half of the rate of the compound data block, one single event sequence can carry more than one application sequence by multiplexing the application sequence into a single event sequence assigned to the compound data block. In this case, each multiplexed application sequence is identified by its DBC (data block count). An example of asynchronous sequence multiplexing is shown in [Figure](#page-44-0) 36.

If the AM824 sequence defines no-data for padding, even an application sequence, which is asynchronous to Transfer Frequency can be adapted to the AM824 sequence. One significant example of this case is the adaptation of a MIDI data stream (application sequence) to a MIDI conformant sequence (AM824 sequence).

Figure 36 – Asynchronous sequence multiplexing

An application that uses this multiplexing shall define MULTIPLEX_NUMBER to be a power of 2, such as 2, 4, 8, 16 and so on. The MULTIPLEX NUMBER is defined in conjunction with the LABEL definition because the place for carrying the MULTIPLEX_NUMBER information is not defined in this standard. This definition will be overridden by a future specification if it defines a method of carrying the MULTIPLEX NUMBER.

The identifier for a multiplexed sequence denoted by MULTIPLEX_INDEX is given by MULTIPLEX_INDEX = mod (DBC, MULTIPLEX_NUMBER).

11.4 Compound data block structure

11.4.1 General

Compound data block is the name for the data block that consists of AM824 data in any combination, if all the AM824 data in the data block specify the same SFC table. (Note that the SFC value in a CIP specifies the entry of the SFC table selected according to the AM824 data type that defines the SFC table.)

Thus the cluster, which is equivalent to a data block in the context of AM824 data, can be referred to as a compound cluster.

Each sequence carried by a compound data block is uniquely identified by the location of events in the compound block.

Figure 37 – Example of compound data block

An example of usage of compound data block is illustrated in [Figure](#page-45-0) 37.

The capital letter, 'B' for example, in the box of AM824 data, represents the box's data type. The small letter, 'b' for example, in the box of AM824 data, denotes no-data for same data type.

DBS (data block size) or CLUSTER_DIMENSION may vary in time. Also, the AM824 data type described in the LABEL field of each event may vary in time.

11.4.2 Compound data structure rule

11.4.2.1 General

[IEC 61883-6:2002](http://dx.doi.org/10.3403/02755535) allows any order of AM824 data type in a compound data block. In order to maintain minimum connectivity, this subclause defines rules for the compound data structure, or, in other words, a rule for the AM824 sequence configuration. A flowchart of the rule is shown in [Figure](#page-45-1) 38. Also, some recommendations for implementation are described.

11.4.2.2 Size rule definition

The size of compound data should be an even number of quadlet.

If the number of quadlet in a sequence required by an application is not an even number, an unspecified sequence (sequence of ancillary no-data with CONTEXT = CF_{16}) should be added to make the number of quadlets in the sequence even. [Figure](#page-45-0) 37 shows a compound data block compliant to this rule where the event denoted by "Z" is interpreted as ancillary no-data with CONTEXT = CF_{16} . As long as the number of quadlets in a sequence is even, any number of unspecified sequences may be added.

11.4.2.3 Order rule definition

An application specifier is either application-specific ancillary data or any common ancillary data except ancillary no-data for non-ancillary data. The content data is any AM824 data other than the application specifier.

A compound data block starts with zero or only one unspecified region followed by zero or one or more specified region(s), see [Figure](#page-46-0) 39 and an example shown in [Figure](#page-47-0) 40. An unspecified region includes only content data. A specified region starts with one or more application specifiers followed by one or more content data before encountering the next application specifier or the end of the compound data block.

A sequence of application specifiers may contain both common ancillary data and applicationspecific ancillary data by multiplexing.

The order of the content data in an unspecified region shall be determined by the following formula.

[IEC 60958](http://dx.doi.org/10.3403/00952384U) Conformant Data < Multi-bit Linear Audio < MIDI Conformant Data < SMPTE Time Code < Sample Count.

Within an unspecified region, the same data type should occupy a contiguous area.

The order inside a specified region is defined by the application specified in the applicationspecific ancillary data. The specified region shall have none or only one common ancillary data or one or more application-specific data for the same application.

Figure 39 – Generic compound data block structure

Figure 40 – Example of unspecified region structure

11.4.2.4 Recommendation: general

Because 2-channel stereo audio is widely accepted, it is highly recommended that for devices, which transmit audio in any format, the first two sequences be linear audio either in [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant or raw audio. The first sequence should be left and the second should be right. If a transmitter is a monaural audio device, it may send the audio in the left channel and silent data in the right, or send the audio in both left and right. It is implementationdependent.

If a transmitter is a multi-channel audio device, it may send down-mixed in 2-channel stereo audio in addition to the multi-channel audio.

11.4.2.5 Recommendation for transmitter

The following recommendations for transmitters apply.

a) The DBS (data block size in quadlet) should be greater than, or equal to, 2. An even number is preferable.

At the top of the data block of mixed audio and music data, the stereo left channel, then the right channel should be transmitted.

- b) In data blocks of multichannel audio data, the first two quadlets should be the main channels corresponding to stereo left and right channel.
- c) Recommendation for stream change method is as follows:

When the contents of stream are changed, it is preferable to insert ancillary no-data or empty packets at the change point of the stream.

The change point of the stream is not a pause of each tune in the CD album, but it implies that at the point, some change of, for example, compression methods, occurs.

The purpose of inserting ancillary no data or empty packets is to prevent losing the end portion of the previous stream and the beginning of the next stream.

The general recommendation method is described as follows.

It is desirable to output ancillary no data with the previous CONTEXT of 10 ms or more following the previous stream.

Afterwards, when the next stream can be recognized beforehand, the insertion of ancillary no-data with the next CONTEXT is recommended.

Otherwise, the insertions of ancillary no data with the next CONTEXT are not needed.

That is, the ancillary no data with the previous CONTEXT can be changed to the following stream directly.

And, when the transmission device does not have the capability of outputting the ancillary no data with the previous CONTEXT and the ancillary no data with next CONTEXT, the transmission device can output MIDI no data or empty packets or stop the stream output.

When the empty packets are output to prevent losing the beginning of the following content, it is preferable to add time stamp information in SYT.

11.4.2.6 Recommendation for receiver

The following recommendations for receivers apply.

- a) Stereo products that receive multichannel streams with DBS >= 2 should reproduce the sound of the first two channels of the data block as stereo left and right channels.
- b) Stereo products which have no non-linear PCM decoder should reproduce no sound (muted) when they receive validity flag = '1' in [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data.

12 AM824 sequence adaptation layers

12.1 Overview

12.1.1 General

The transport mechanism using CIP may be used as an alternative transport layer for an existing data transmission protocol such as [IEC 60958](http://dx.doi.org/10.3403/00952384U) and MIDI.

This adaptation layer definition defines only one-to-one mapping between an application data structure and an AM824 data structure and a procedure for transporting the application data only with a constant time shift.

The definition of the adaptation to CIP can be described and maintained by either organization responsible for the adaptation.

The adaptation layer definition described in this standard provides only an alternative transport. The meaning of the data carried by the transport should be given in the original specification. Also, the transmission rate should be identical to that which is originally specified when the "non-identical to sampling frequency" indication flag is off.

The adaptation layer definition falls into two categories. One is generic that can be used in applications and does not define application-specific ancillary data. Another is applicationspecific that defines the structure of the compound data block and application-specific ancillary data.

12.1.2 [IEC 60958](http://dx.doi.org/10.3403/00952384U) bitstream

12.1.2.1 General

All the information defined in [IEC 60958](http://dx.doi.org/10.3403/00952384U) is mapped into this data format. Application, which uses [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data, shall follow the [IEC 60958](http://dx.doi.org/10.3403/00952384U) standard.

12.1.2.2 Sampling frequency in [IEC 60958-3](http://dx.doi.org/10.3403/02859696U)

In [IEC 60958-3](http://dx.doi.org/10.3403/02859696U), six new sampling frequencies of 22,05 kHz, 24 kHz, 88,2 kHz, 96 kHz, 176,4 kHz and 192 kHz are defined by bit 24 to 27 "sampling frequency" of channel status, as shown in [Table](#page-49-0) 25. All other combinations are reserved and shall not be used until further defined.

Table 25 – Sampling frequency in [IEC 60958-3](http://dx.doi.org/10.3403/02859696U)

12.1.2.3 Original sampling frequency

Bit 36 to 39 is defined as the "original sampling frequency", as shown in [Table](#page-49-1) 26.

Table 26 – Original sampling frequency

12.1.2.4 Relation of sampling frequency and original sampling frequency

The sampling frequencies in [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) have relations of multiple of integer number as follows.

- 32 kHz line: (8 kHz, 16 kHz), 32 kHz
- 44,1 kHz line: (11,025 kHz), 22,05 kHz, 44,1 kHz, 88,2 kHz, 176,4 kHz
- 48 kHz line: (12 kHz), 24 kHz, 48 kHz, 96 kHz, 192 kHz

Sampling frequencies in parenthesis are defined only in original sampling frequency. Original sampling frequency is recorded in disc or transmitted by broadcasting and supplied from source devices, e.g. players or tuners.

12.1.2.5 Up or down sampling ratio

Original sampling frequency can be up-sampled or down-sampled. When the up- or downsampling ratio is defined, the relation of sampling frequency and original sampling frequency is expressed using the following formula.

Sampling frequency = original sampling frequency * up or down sampling ratio (10)

Examples of sampling ratio are shown in [Table](#page-50-0) 27, [Table](#page-50-1) 28 and [Table](#page-50-2) 29.

Original sampling	Sampling frequency
frequency	32 kHz
8 kHz	
16 kHz	
32 kHz	

Table 27 – Up or down sampling ratio of 32 kHz line

Table 28 – Up or down sampling ratio of 44,1 kHz line

Table 29 – Up or down sampling ratio of 48 kHz line

12.1.2.6 Clock accuracy in [IEC 60958-3](http://dx.doi.org/10.3403/02859696U)

In [IEC 60958-3](http://dx.doi.org/10.3403/02859696U), "11" in bit 28-29 of channel status is defined as "interface frame rate not matched to sampling frequency", as shown in [Table](#page-51-0) 30.

Table 30 – Clock accuracy in [IEC 60958-3](http://dx.doi.org/10.3403/02859696U)

12.1.2.7 High-speed transmission ratio and interface frame rate

High-speed transmission can be executed over [IEC 60958](http://dx.doi.org/10.3403/00952384U) digital audio interface. Original sampling frequency, high-speed transmission ratio and interface frame rate is expressed using the following formula.

> Interface frame rate = original sampling frequency \times up or down sampling ratio \times high-speed transmission ratio (11)

Clock accuracy "11" in [Table](#page-51-0) 30 means that high-speed transmission is executed. When clock accuracy is "11", there are two cases as follows:

- a) the sampling frequency is equal to the original sampling frequency;
- b) the sampling frequency is not equal to the original sampling frequency.

The former case means that there is no up or down sampling process and the latter case means that there is an up or down sampling process. When clock accuracy is "11", the interface frame rate may be different from the sampling frequency.

Clock accuracy "00", "01" or "10" means that there is no high-speed transmission. When clock accuracy is "00", "01" or "10", there are two cases as follows:

- 1) the sampling frequency is equal to the original sampling frequency;
- 2) the sampling frequency is not equal to the original sampling frequency.

The latter case means there is an up or down sampling process and the former case means that there is neither up or down sampling nor high-speed transmission. When the clock accuracy is "00", "01" or "10", the interface frame rate is the same as sampling frequency.

These cases are illustrated in [Table](#page-51-1) 31.

Clock accuracy	Original sampling frequency	Sampling frequency	Interface frame rate	Case
11	Original sampling frequency	Not equal to original sampling frequency	Not equal to sampling frequency	High-speed transmission and up or down sampling
11	Original sampling frequency	Equal to original sampling frequency	Not equal to sampling frequency	High-speed transmission
00, 01, 10	Original sampling frequency	Not equal to original sampling frequency	Equal to sampling frequency	Up or down sampling
00, 01, 10	Original sampling frequency	Equal to original sampling frequency	Equal to sampling frequency	Original

Table 31 – Cases

In [Table](#page-52-0) 32, some examples of cases are described.

Source device condition				Interface condition				
Original sampling frequency	Up or down sampling ratio	Sampling frequency	High-speed transmission ratio	Interface frame rate	Clock accuracy	Original sampling frequency	Sampling frequency	
					Bit 28, 29	Bit 36-39	Bit 24-27	
	$\overline{2}$	88,2 kHz	$\mathbf{1}$	88,2 kHz	00,01,10		0001	
			1	44,1 kHz	00,01,10	1111		
44,1 kHz	1	44,1 kHz	$\overline{2}$	88,2 kHz	11		0000	
			4	176,4 kHz	11			
			$\mathbf{1}$	96 kHz	00,01,10		0101	
1	96 kHz	2	192 kHz	11				
96 kHz	1/2			1	48 kHz	00,01,10	1010	
		48 kHz	2	96 kHz	11		0100	
			4	192 kHz	11			
	$\mathbf{1}$	192 kHz	1	192 kHz	00,01,10		0111	
			1	96 kHz	00,01,10			
	1/2	96 kHz		2	192 kHz	11		0101
192 kHz 1/4		1	48 kHz	00,01,10	1000			
		48 kHz	2	96 kHz	11		0100	
			4	192 kHz	11			
If the interface frame rate is equal to the original sampling frequency, there may be an up or down NOTE sampling process and a high-speed transmission process.								

Table 32 – Examples

12.1.2.8 N-flag definition

N-flag is used with AM824 data in general (see [10.1,](#page-40-2) [10.4](#page-42-0) and AV/C command set for rate control of isochronous data flow 1.0).

When N-flag = 1, RATE CONTROL command with BASE CONFIGURE subfunction can execute high-speed transmission of AM824 data over IEEE 1394 and RATE CONTROL command with FLOW CONTROL subfunction can execute flow control of AM824 data. This subclause describes the relation between channel status coding in [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) and N-flag, SFC and SYT-INTERVAL in [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data, as shown in [Table](#page-54-0) 33.

With the introduction of [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) and command-based RATE CONTROL, the following cases may happen:

- a) real-time transmission of 96 kHz (or 192 kHz) original sampling PCM signal over [IEC 60958](http://dx.doi.org/10.3403/00952384U);
- b) real-time transmission of up-sampled 48 kHz original sampling PCM signal by 96 kHz (or 192 kHz) sampling frequency over [IEC 60958](http://dx.doi.org/10.3403/00952384U);
- c) high-speed transmission of 48 kHz original sampling PCM signal with 96 kHz (or 192 kHz) sampling frequency over [IEC 60958;](http://dx.doi.org/10.3403/00952384U)
- d) double (or four times) high-speed transmission of 48 kHz original sampling PCM signal with RATE CONTROL command with BASE CONFIGURE subfunction over the IEEE 1394 bus.

When [IEC 60958](http://dx.doi.org/10.3403/00952384U) signals of a), b) and c) are transmitted over IEEE 1394 with [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant mode, some mechanisms are necessary to distinguish signals of a), b), c) and d) on isochronous mode.

For cases of a), b) and c), [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) specifies codes in clock accuracy (see [Table](#page-51-0) 30) and original sampling frequency (see [Table](#page-49-1) 26) in channel status. For d), N-flag is set to "1".

When N-flag = 0, values of SYT_INTERVAL and Nominal_Sampling_Frequency are described in [Table](#page-39-0) 20 for FDF = 0000 $0xxx_2$. When N-flag = 1, values of SYT_INTERVAL and Nominal_Sampling_Frequency are described in [Table](#page-39-0) 20 for FDF = $\overline{0000}$ 1xxx₂. For IFC 60958 conformant data, the following rules for SYT INTERVAL and rules for SYT_INTERVAL Nominal_Sampling_Frequency are applied.

- 1) When N-flag = 0, value of Nominal_Sampling_Frequency for [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data is set according to interface frame rate and value of SYT_INTERVAL is set corresponding to the Nominal Sampling Frequency in [Table](#page-39-1) 21.
- 2) When N-flag = 1, value of Nominal_Sampling_Frequency for [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data is set according to sampling frequency coded in Bit 24-27 of channel status of [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) format.
- 3) When N-flag = 1 and RATE CONTROL command with BASE CONFIGURE subfunction is executed, value of SYT_INTERVAL is set to *n* multiplexed by SYT_INTERVAL value corresponding to the Nominal_Sampling_Frequency in [Table](#page-43-0) 24.
- 4) When N-flag = 1 and RATE CONTROL command with BASE CONFIGURE subfunction is executed, clock accuracy, bit 28-29 of channel status, of [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) is set to '11'.
- 5) When N-flag = 1 and RATE CONTROL command with FLOW CONTROL Subfunction is executed, value of SYT_INTERVAL is set corresponding to the Nominal_Sampling*_* Frequency in [Table](#page-42-2) 23.
- 6) When N-flag = 1 and RATE CONTROL command with FLOW CONTROL subfunction is executed, clock accuracy, bit 28-29 of channel status, of [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) is set to '00', '01' or '10'.

	IEC 60958-3					A/M protocol		
	Bit 36-39	Bit 24-27	Bit 28 29	Interface				
	Original sampling frequency	Sampling frequency	Clock accuracy	frame rate	N- flag	SFC	SYT- INTERVAL	
Case a)	96 kHz	96 kHz		96 kHz				
Original	(192 kHz)	(192 kHz)	00,01,10	(192 kHz)	$\overline{}$			
IEC 60958-3								
Case b)		96 kHz		96 kHz				
Up sampling	48 kHz	(192 kHz)	00,01,10	(192 kHz)	—			
IEC 60958-3								
Case c)				96 kHz				
High-speed	48 kHz	48 kHz	11	(192 kHz)	$\overline{}$			
IEC 60958-3								
Case a)	96 kHz	96 kHz		96 kHz		96 kHz	16	
Original	(192 kHz)	(192 kHz)	00,01,10	(192 kHz)	Ω	(192 kHz)	(32)	
with A/M Protocol								
Case b)		96 kHz		96 kHz		96 kHz	16	
Up sampling	48 kHz	(192 kHz)	00,01,10	(192 kHz)	$\mathbf 0$	(192 kHz)	(32)	
with A/M Protocol								
Case c)				96 kHz		96 kHz	16	
High-speed	48 kHz	48 kHz	11	(192 kHz)	$\mathbf 0$	(192 kHz)	(32)	
with A/M Protocol								
Case d)				96 kHz			$8 * n$	
Rate control	48 kHz	48 kHz	11	$(192$ kHz)	$\mathbf{1}$	48 kHz	$(n = 2, 4)$	
with A/M Protocol								
FLOW	96 kHz	96 kHz		96 kHz		96 kHz	16	
CONTROL	(192 kHz)	(192 kHz)	00,01,10	(192 kHz)	$\mathbf{1}$	(192 kHz)	(32)	
with A/M Protocol								

Table 33 – Relation of values in [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) and A/M protocol

12.1.3 One-bit audio

12.1.3.1 General

In this subclause, the format of one bit audio is described.

12.1.3.2 One-bit audio (plain)

The data of the one-bit audio (LABEL = 50_{16} - 51_{16}) has one-bit length data stream and can be directly played back through the analogue low pass filter bit by bit (MSB first). The data stream is packed in 24-bit data fields of an AM824 quadlet, as shown in [Figure](#page-55-0) 41, with MSB first per audio channel.

The sampling frequency of the one-bit audio (LABEL = 50_{16} , 51_{16} , 58_{16}) is defined in [Table](#page-55-1) 34 with its own SFC table.

Table 34 – Sampling frequency definition of one-bit audio

The TRANSFER_DELAY for blocking transmission, in the case of DEFAULT_TRANSFER_ DELAY = 479,17 μ s = (354,17 + 125) μ s, corresponds to [Table](#page-55-2) 34, as given in Table 35.

Table 35 – TRANSFER_DELAY for blocking transmission in the case of the one-bit audio

The one-bit audio (LABEL = 50_{16} – 51_{16}) can transmit multichannel cluster. Each AM824 quadlet carries the data for one channel of the cluster. Two AM824 LABELs are used to indicate the start and continuation of the data in the cluster.

Figure 41 – Generic one-bit audio quadlet

The channel number shall start with No.1 and be sequential. An example is given in [Figure](#page-56-0) 42.

: :

Data for other channels in the cluster **:**

:

Figure 42 – Generic one-bit audio quadlet sequence

12.1.3.3 One-bit audio (encoded) – DST

The data of the one-bit audio (encoded) is the encoded data stream.

DST (direct stream transfer) is the loss-less coding technique used for one-bit audio in SACD and is defined in super audio CD system description version 1.2.

The encoded data stream is packed in 24-bit data fields of AM824 data, as shown in [Figure](#page-56-1) 43, with MSB first.

For decoding the stream, SACD ancillary data is needed. DST supports multichannel one-bit audio and carries each data stream in one mixed stream.

DST encodes the one-bit audio data stream frame by frame. The frame is defined in super audio CD system description version 1.2.

The sampling frequency of the DST is defined in [Table](#page-55-1) 34 with its own SFC table.

Figure 43 – One-bit audio DST encoded quadlet

12.1.3.4 High-speed transfer for one-bit audio

As far as one-bit audio (LABEL = 50_{16} , 51_{16} , 58_{16}) is concerned, the transfer frequency and SYT_INTERVAL for the high-speed AM824-data transfer are defined depending on the speed shown in [Table](#page-57-0) 36 if the N-flag in the FDF is 1. In Table 36, an integer value of n (>1) indicates the number of times faster than normal speed.

Table 36 – SFC definition of one-bit audio for high-speed AM824 data transfer

The DBS of an event is independent of the transfer speed.

12.1.4 Non-linear audio data stream

Any non-linear audio data carried by an [IEC 61937](http://dx.doi.org/10.3403/02223327U) bit-stream can be transmitted by using the [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data sequence.

12.1.5 MIDI data stream

Any modification or enhancement is prohibited on this adaptation layer, although increase of transmission rate, for instance, can be easily done. The specification that uses this adaptation layer is given in MMA/AMEI RP-027.

This standard restricts the packetization of MIDI data stream so that a single MIDI conformant sequence can carry multiple MIDI data streams by multiplexing. The MIDI conformant data defines MULTIPLEX_NUMBER = 8.

NOTE The default MULTIPLEX NUMBER for MIDI conformant AM824 types may be incompatible with some applications conforming to [IEC 61883-6:2002](http://dx.doi.org/10.3403/02755535).

Figure 44 – Multiplexing of MIDI data streams

[Figure](#page-57-1) 44 shows how two MIDI data streams, which should flow in different MIDI cables, are multiplexed in a single MIDI conformant sequence with an audio stream. [Figure 44](#page-57-1) gives only the sequence multiplexing scheme. The parameters of this example such as the number of multiplexed sequences and the audio sampling rate were chosen so that the figure be

readable. Consequently, not all the parameters are valid for this standard (including its previous editions).

12.1.6 SMPTE time code and sample count

SMPTE time code and sample count transmission are defined in TA 1999024 SMPTE Time Code and Sample Count Transmission Protocol Version 1.0.

12.1.7 High-precision and double-precision multi-bit linear audio

12.1.7.1 High-precision specific ancillary sata

This subclause specifies private header data that are carried by high-precision specific ancillary data.

These ancillary data are transmitted at every data block. The data format is shown in [Figure](#page-58-0) 45. [Table](#page-58-1) 37 and [Table](#page-58-2) 38 provide channel and accuracy definitions, respectively.

Figure 45 – High-precision first ancillary data

Table 37 – Channel definition

Table 38 – Accuracy definition

With the combination of accuracy and Num. (slot number), any PCM audio data with sample word length up to 192 bit can be transmitted by high-precision multi-bit linear audio. There is a wide redundancy, for example, a 64-bit sample word can be transmitted with 3 slots of a 24-bit slot (acc = $10₂$) or 4 slots of a 16-bit slot (acc = $00₂$). To eliminate hardware complexity on the decoder side, the following implementation rules are strongly recommended:

- a) sample word should be limited to 32, 40, 48, 64, 80, 96, 128, 160 and 192 bit length;
- b) number of slots should be limited to 2, 4 and 8.

The accuracy for the above stated sample words should be as specified in [Table](#page-59-0) 39.

Table 39 – Recommended rules

When a source device sends its own data or auxiliary information to the sink device in highprecision mode, its original data and/or ancillary data can be transmitted between the highprecision first ancillary data and high-precision multi-bit linear audio data.

For example, [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data can be transmitted between the high-precision first ancillary data and high-precision multi-bit linear audio data, see [Figure](#page-59-1) 46. For other applications, common and application-specific ancillary data can be transmitted between the high-precision first ancillary data and high-precision multi-bit linear audio data, see [Figure](#page-60-0) 47. Refer to [8.2.9.2.1](#page-34-0) and [8.2.10](#page-36-3) and each application section.

Figure 46 – [IEC 60958](http://dx.doi.org/10.3403/00952384U) conformant data with high-precision data

Figure 47 – Common and application-specific ancillary data with high-precision data

These ancillary data are optional and their definition is reserved. [Figure](#page-60-1) 48 shows highprecision channel assignment ancillary data and [Table](#page-60-2) 40 provides a channel assignment definition.

Figure 48 – High-precision channel assignment ancillary data

Channel assignment	Description
0000 0000 0000 0000,	
0000 0000 0000 0001,	
0000 0000 0000 0010,	
.	Reserved
1111 1111 1111 1110,	
1111 1111 1111 1111 ₂	

Table 40 – Channel assignment definition

12.1.7.2 Example of high-precision stream

[Figure](#page-61-0) 49 shows a 2-channel 128-bit sample-word high-precision stream carried over the serial bus. Here, the lower 8 bit are set to "0". High-precision ancillary data are immediately followed by the data block.

Figure 49 – Example of high-precision data

12.1.7.3 Example of double-precision stream

[Figure](#page-62-0) 50 shows a 6-channel 48-bit sample-word double-precision stream carried over IEEE 1394.

Note that double precision uses the LABELs from 60_{16} to 61_{16} .

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Figure 50 – Example of double-precision data

12.1.7.4 Example of double-precision compound stream

[Figure](#page-63-1) 51 shows a 4-channel 48-bit sample-word double-precision compound stream carried over IEEE 1394.

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Figure 51 – Example of double-precision compound data

12.2 DVD-audio

12.2.1 General

The compound data for DVD-Audio consists of multi-bit linear audio data, common ancillary and DVD-Audio specific ancillary data.

12.2.2 DVD-Audio specific ancillary data

12.2.2.1 General

This subclause specifies private header data that are carried by DVD-Audio specific ancillary data, as shown in [Table](#page-64-0) 41.

LABEL	SUB LABEL	Description
$D0_{16}$	01_{16}	Data transmitted at every data block
	02_{16}	Data transmitted at starting-point
	CO ₁₆	Audio CCI
	$C1_{16}$	ISRC

Table 41 – DVD-Audio specific ancillary data

12.2.2.2 Data transmitted at starting-point

These ancillary data, shown in [Figure](#page-64-1) 52, are used at the starting-point of audio data when performing play start or search for a track number. [Table](#page-64-2) 42 provides definitions of Fs2, Multichannel type, channel assignment and table parity.

Figure 52 – Data transmitted at data starting-point

Data	Bit	Description
Fs2	4	Sampling frequency group2
Multi-channel type	4	Fs, bit combination table
Channel assignment	5	Channel combination of group1 and 2
Table parity		Table parity of audio data

Table 42 – Data transmitted at starting-point

12.2.2.3 Data transmitted at every data block

These ancillary data, shown in [Figure](#page-64-3) 53, are transmitted at every data block. [Table](#page-65-0) 43 provides definitions of dynamic range control, down mix code, emphasis flag, down mix mode and down mix code validity.

Figure 53 – Data transmitted at every data block

Table 43 – Data transmitted at every data block

12.2.3 Data for CCI

[Figure](#page-65-1) 54 shows ancillary data for CCI.

SUB LABEL $CO₁₆$ is for CCI.

Figure 54 – Ancillary data for CCI

NOTE Audio CCI signifies copy control information for audio.

12.2.4 Data for ISRC

[Figure](#page-65-2) 55 shows ancillary data for ISRC.

SUB LABEL $C1_{16}$ is for ISRC.

Figure 55 – Ancillary data for ISRC

12.2.5 Example of DVD-Audio stream

[Figure](#page-66-0) 56 illustrates a basic data block of DVD-Audio stream carried over IEEE 1394 in the case of six channels.

$D0_{16}$	01_{16}	Ancillary Data
$D0_{16}$	02_{16}	Ancillary Data
48 ₁₆		Data channel 1
48_{16}		Data channel 2
48_{16}		Data channel 3
4816		Data channel 4
4816		Data channel 5
48 ₁₆		Data channel 6

Figure 56 – Basic data block of DVD-Audio stream

Data on the disc is organized into a series of blocks. The data for each channel is packed into one block. Each data block should be ordered by increasing channel number. The data block immediately follows DVD-Audio ancillary data. The first ancillary data is "the data transmitted at every data block," and the second ancillary data is "the data transmitted at the data starting point" or "table parity" or "DMCT (down mix coefficient table)" or something similar.

[Figure](#page-67-1) 57 illustrates an example of DVD-Audio data stream that carries scaleable contents of DVD-Audio. In this case, the sampling frequency and sample word length may be different between front channels and rear channels, and, in the second data block, previous data hold of ASI2 of DVD-Audio is used.

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Figure 57 – Example of DVD-Audio data

12.3 SACD definition

12.3.1 General

———————

The data block for SACD consists of one-bit audio data, common ancillary and SACD-specific ancillary data.

12.3.2 SACD ancillary data

The SACD player transmits SACD ancillary data at the starting-point of every frame. The frame is defined in [[1](#page-67-2)2]¹. The SACD ancillary data is shown in [Figure](#page-68-0) 58, and [Table](#page-68-1) 44 provides definitions of validity flag, Track_Attribute, Ch_Bit_n and Loudspeaker_Config.

¹ Numbers in square brackets refer to the Bibliography.

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Figure 58 – SACD ancillary data

Table 44 – Data information

Data Bit Description Validity flag 1 Valid or not valid Track_Attribute 14 Copy control information Ch_Bit_n 3 Number of channels Loudspeaker_Config | 5 | Loudspeaker set-up

The validity flag shows the validity of the data within the frame.

If a disc read error occurs, the SACD player shall replace the error data with safe data, such as a mute signal, and set the validity flag to $1₂$, as shown in [Table](#page-68-2) 45.

Rsv is the reserved area and the default value is $000₂$.

The Track Attribute shows copy control information dedicated to super audio CD, and is defined in [1[2](#page-68-3)] ². This information shall be copied from the super audio CD track by track.

The Ch_Bit_n shows the total number of channels, and is defined in [12]. This information shall be copied from the super audio CD frame by frame.

The Loudspeaker Config shows the loudspeaker set-up, and is defined in [12]. This information shall be copied from the super audio CD track by track.

12.3.3 SACD supplementary data

———————

SACD supplementary data is a synchronized stream along with the audio data from the SACD, see [Figure](#page-69-0) 59. It has several data lengths as defined in [12]. Audio data and supplementary data are synchronized on a frame-by-frame basis.

For decoding the stream, SACD ancillary data is needed.

² Numbers in square brackets refer to the Bibliography.

Figure 59 – SACD supplementary data

12.3.4 Multi-bit linear audio data

DVD-Audio data use the LABEL from 48_{16} to $4F_{16}$ of multi-bit linear audio and uses ASI2 for scaleable contents, as shown in [Table](#page-69-1) 46.

Value	Description
00 ₂	24 bit
01,	20 bit
10 ₂	16 bit
11 ₂	Previous sample word data hold

Table 46 – ASI2 definition for DVD-Audio

12.3.5 SACD Track_Mode&Flags data

SACD Track_Mode&Flags data consists of Track_Mode (1 B) and track flags (1 B) defined in [12]. The relationship between ACD Track Mode&Flags,Track Mode and Track Flags is described in [Figure](#page-69-2) 60.

Figure 60 – SACD Track_Mode&Flags data

12.3.6 SACD Track_Copy_Management data

SACD Track_Copy_Management data consists of three AM824 data quadlet and shows the Track Copy Management defined in [12]. The data of the Track Copy Management (6 B) is divided into three data fields (Part 1, 2, 3) of the AM824 quadlet (AM824 LABEL=D1₁₆: SUB LABEL=10₁₆, 11₁₆, 12₁₆) in sequence, see [Figure](#page-69-3) 61.

12.3.7 Example of SACD streams

[Figure](#page-70-0) 62 illustrates a typical multi-channel plain one-bit audio stream carried over the 1394 bus from SACD for the case where the value of SFC in FDF is $001₂$. The data on the disc is

organized into a series of frames, with 75 frames for each second of audio. Each frame contains a total of (1 568 \times 3) B of audio cluster data per channel. Quadlets in a data block are organized according to the order rule, so that the order is ancillary data first, multichannel cluster data next, and an ancillary no data with CONTEXT = CF_{16} last.

The SACD ancillary data starts and is followed by the first group of multi-channel cluster data. In this example, the first quadlet contains the ancillary data for the whole of frame #0. If, for example, there is a disc error, the SACD player sets the validity flag in the ancillary data for this frame (frame #0) which remains valid until the next SACD ancillary data (frame #1). This also applies to the Track_Attribute, Ch_Bit_n and Loudspeaker_Config contained in the ancillary data for frame #0.

Figure 62 – Example of SACD stream in the case of six channels

In the example of [Figure](#page-70-0) 62, there are six channels in the multi-channel cluster, so an ancillary No-Data with CONTEXT = CF is added to the last of the cluster data so that the total numbers of quadlets in the block is kept even, and therefore DBS = 8. The SACD supplementary data is transmitted at the same location as the SACD ancillary data (after the SACD ancillary data has already been transmitted). After all the SACD supplementary data has been transmitted, an ancillary no data or other ancillary data quadlet may be put in the same location.

[Figure](#page-71-1) 63 shows five channel cases. Here, ancillary no data with CONTEXT = CF is not required, and the DBS = 6.

Figure 63 – Example of SACD stream in the case of five channels

12.4 Blu-ray Disc

12.4.1 General

The compound data for Blu-ray Disc consists of multi-bit linear audio data, common ancillary and Blu-ray Disc specific ancillary data.

12.4.2 Structure of sample word for audio transmission

[Figure](#page-72-0) 64 shows basic data blocks of Blu-ray Disc. There are eight sample words in one audio sample.

Figure 64 – Basic data blocks of Blu-ray Disc

Channel layout is fixed.

The transmitter shall set 000000₁₆ on the audio data of MBLA data in the case of a non existing channel.

Valid combinations of the sample word are determined by the permitted channel allocations, as defined in [15].

12.4.3 Multi-bit linear audio data

Blu-ray Disc data use the LABEL from 48₁₆ to 4A₁₆ of multi-bit linear audio. [Table](#page-72-0) 47 and [Table](#page-72-1) 48 provide definitions of ASI1 and ASI2, respectively.

Table 47 – ASI1 definition for Blu-ray Disc

Table 48 – ASI2 definition for Blu-ray Disc

12.4.4 Blu-ray Disc specific ancillary data

This subclause specifies private header data that are carried by Blu-ray Disc specific ancillary data and as shown in [Table](#page-72-2) 49.

Table 49 – Blu-ray Disc specific ancillary data

LABEL	SUB LABEL	Description
$D3_{16}$	01_{16}	Data transmitted at every data block
	$\mathsf{CO}_{4\,c}$	CCI

The transmission device shall execute the stream change method if the ancillary data is changed except when SUB LABEL is $CO₁₆$.

12.4.5 Data transmitted at every data block

This following ancillary data is transmitted at every data block, see [Figure](#page-73-0) 65 and [Table](#page-73-1) 50.

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Figure 65 – Data transmitted at every data block

Table 50 – Data transmitted at every data block

The L channel shows whether L (left) channel data exist or not, as shown in [Table](#page-73-2) 51.

Table 51 – L channel definition

The R channel shows whether R (right) channel data exist or not, as shown in [Table](#page-73-3) 52.

Table 52 – R channel definition

The lfe channel shows whether lfe (low frequency effects) channel data is existed or not, as shown in [Table](#page-74-0) 53.

Table 53 – lfe channel definition

The C channel shows whether C (Centre) channel data is existed or not, as shown in [Table](#page-74-1) 54.

Table 54 – C channel definition

Value	Description
	C (centre) channel data do not exist
	C (centre) channel data exist

The LS channel shows whether LS (left surround) channel data exist or not, as shown in [Table](#page-74-2) 55.

Table 55 – LS channel definition

The RS channel shows whether RS (right surround) channel data exist or not, as shown in [Table](#page-74-3) 56.

Table 56 – RS channel definition

The Rls channel shows whether Rls (rear surround left) channel data exist or not, as shown in [Table](#page-74-4) 57.

Table 57 – Rls channel definition

The Rrs channel shows whether Rrs (rear surround right) channel data exist or not, as shown in [Table](#page-74-5) 58.

Table 58 – Rrs channel definition

The L/R ch identifier shows whether L (left)/R (right) channel data is L/R signal (stereo) or M1 (mono) signal or Lo (left output)/Ro (right output) signal or Lt (left total)/Rt (right total) signal.

Note that the sink device shall decrease the M1 (mono) signal to a −3 dB level if the sink device outputs M1 (mono) signal in the L (left)/R (right) channel, as shown in [Table](#page-75-0) 59.

Table 59 – L/R ch identifier definition

The C ch identifier shows whether the C (Centre) channel data is the C signal or M1 (mono) signal as shown in [Table](#page-75-1) 60.

Table 60 – C ch identifier definition

The LS/RS ch identifier shows whether LS (left surround)/RS (right surround) channel data is the LS/RS signal or S (surround) signal.

Note that the sink device shall decrease the S (surround) signal to a −3 dB level if the sink device outputs S (surround) signal in the LS (left surround)/RS (right surround) channel as shown in [Table](#page-75-2) 61.

Table 61 – LS/RS ch identifier definition

12.4.6 Data for CCI

[Figure](#page-76-0) 66 shows ancillary data for CCI, and a data field definition is provided in [Table](#page-76-1) 62.

SUB LABEL $CO₁₆$ is for CCI.

Figure 66 – Ancillary data for CCI

Table 62 – Data transmitted at every data block

Value	Bit
1) reserved	
2) Retention_Move_Mode	1
3) Retention_State	3
4) EPN	
5) CCI	2
6) Digital_Only_Token	
7) Analog Sunset Token	1
8) Image Constraint Token	
9) APS	2

NOTE Ancillary data for CCI is for copy control information and is defined in [15].

These data shall be transmitted at least once during 100 ms period.

12.4.7 Example of Blu-ray Disc stream

[Figure](#page-76-2) 67 illustrates a basic data block of the Blu-ray Disc stream carried over the 1394 Bus in the case of eight channels.

[Figure](#page-77-0) 68 shows a one-channel example.

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a) Example 1

b) Example 1

c) Example 3

Figure 68 – Examples of Blu-ray Disc stream of one channel

The channel structure shown in example 3 of [Figure](#page-77-0) 68 appears when only L and R are transmitted.

[Figure](#page-78-0) 69 shows two-channel examples.

$D3_{16}$	01_{16}	Ancillary Data
$D3_{16}$	CO ₁₆	Ancillary Data
48_{16}		Left
48_{16}		Right
48_{16}		Set to 000000h

a) Example 1

b) Example 2

Figure 69 – Example of Blu-ray Disc stream of two channels

[Figure](#page-79-0) 70 shows a three-channel example.

Figure 70 – Example of Blu-ray Disc stream of three channels (3/0)

[Figure](#page-79-1) 71 shows another three -channel structure.

Figure 71 – Example of Blu-ray Disc stream three channels (2/1)

[Figure](#page-80-0) 72 shows a four-channel structure.

Figure 72 – Example of Blu-ray Disc stream of four channels (2/2)

12.5 Multi-bit linear audio (MBLA)

12.5.1 General

The compound data for MBLA (Multi-Bit Linear Audio) consists of multi-bit linear audio data, common ancillary and multi-bit linear audio specific ancillary data.

12.5.2 Structure of sample word for audio transmission

There are two kinds of channel structures in which this condition is defined:

- a) the fixed channels' structure; and
- b) the variable channels' structure.

12.5.3 Fixed channels' structure of sample word for audio transmission

MBLA specifies thirty-two channels. Thirty-two channels are transmitted by four isochronous channels.

There are eight sample words in one audio sample at each isochronous channel.

[Figure](#page-82-0) 73 shows channels included in each group and channel order.

Group 2

Group 3

Group 4

Figure 73 – Basic data block of the fixed channels' structure

The channel structure layout is fixed, as shown in [Figure](#page-82-0) 73.

The transmitter shall set to $000000₁₆$ on the audio data of the MBLA data in the case of a non existing channel. The group does not need to transmit if all channels do not exist in the group. It saves the isochronous channel resource for the group.

12.5.4 Variable channels' structure of sample word for audio transmission

MBLA specifies thirty-two channels. Thirty-two channels are transmitted by one isochronous channel.

[Figure](#page-84-0) 74 shows the channel order.

Figure 74 – Basic data block of the variable channels' structure

The non existing channel shall not transmit MBLA data.

The channel order is reduced.

12.5.5 MBLA data

MBLA data use the LABEL from 40_{16} to 42_{16} of MBLA.

12.5.6 MBLA specific ancillary data

This subclause specifies private header data that are carried by MBLA specific ancillary data, as shown in [Table](#page-84-1) 63.

The transmission device shall execute stream change method if the ancillary data is changed except when SUB LABEL is $CO₁₆$.

12.5.7 Data transmitted at every data block of Group 1 for the fixed channels' structure

These ancillary data, as shown in [Figure](#page-85-0) 75 and [Table](#page-85-1) 64, are transmitted at every data block of Group 1 for the fixed channels' structure.

Figure 75 – Data transmitted at every data block of Group 1 for the fixed channels' structure

Table 64 – Data transmitted at every data block of Group 1 for the fixed channels' structure

The emphasis flag shows whether de-emphasis is required for the sink device or not, as shown in [Table](#page-85-2) 65.

Table 65 – Emphasis flag definition

The FL channel shows whether FL (front left) channel data exist or not, as shown in [Table](#page-85-3) 66.

Table 66 – FL channel definition

Value	Description
	FL (front left) channel data does not exist
	FL (front left) channel data exist

The FR channel shows whether FR (front right) channel data exist or not, as shown in [Table](#page-86-0) 67.

The LFE1 channel shows whether LFE1 (low frequency effects-1) channel data exist or not, as shown in [Table](#page-86-1) 68.

Table 68 – LFE1 channel definition

Value	Description
Uλ	LFE1 (low frequency effects-1) channel data do not exist
	LFE1 (low frequency effects-1) channel data exist

The FC channel shows whether FC (front centre) channel data exist or not, as shown in [Table](#page-86-2) 69.

Table 69 – FC channel definition

The LS channel shows whether LS (left surround) channel data is existed or not, as shown in [Table](#page-86-3) 70.

Table 70 – LS channel definition

The RS channel shows whether RS (right surround) channel data exist or not, as shown in [Table](#page-86-4) 71.

Table 71 – RS channel definition

The BL channel shows whether BL (back left) channel data exist or not, as shown in [Table](#page-86-5) 72.

Table 72 – BL channel definition

The BR channel shows whether BR (back right) channel data exist or not, as shown in [Table](#page-87-0) 73.

The FL/FR ch identifier shows whether FL (front left)/FR (front right) channel data is FL/FR signal (stereo) or M1 (mono) signal or Lo (left output)/Ro (right output) signal or Lt (left total)/Rt (right total) signal.

Note that the sink device shall decrease the M1 (mono) signal to a −3 dB level if the sink device outputs M1 (mono) signal in the L (left)/R (right) channel, as shown in [Table](#page-87-1) 74.

Value	Description
00 ₂	FL (front left)/FR (front right) channel data is a L/R (stereo) signal
01,	FL (front left)/FR (front right) channel data is a M1 (mono) signal
	FL (front left) channel and FR (front right) channel data are the same
10 $\rm _2$	FL (front left)/FR (front right) channel data is Lo (left output)/Ro (right output) signal
11,	FL (front left)/FR (front right) channel data is a Lt (left total)/Rt (right total) signal

Table 74 – FL/FR ch identifier definition

The FC ch identifier shows whether FC (front centre) channel data is a FC signal or a M1 (mono) signal, as shown in [Table](#page-87-2) 75.

Table 75 – FC ch identifier definition

Value	Description
	FC (front centre) channel data is a FC signal
	FC (front centre) channel data is a M1 (mono) signal

12.5.8 Data transmitted at every data block of Group 2 for the fixed channels' structure

These ancillary data, as shown in [Figure](#page-87-3) 76 and [Table](#page-88-0) 76, are transmitted at every data block of Group 2 for the fixed channels' structure.

Figure 76 – Data transmitted at every data of Group 2 for the fixed channels' structure

Table 76 – Data transmitted at every data of Group 2 for the fixed channels' structure

The Emphasis flag shows whether de-emphasis is required for the sink device or not, as shown in [Table](#page-88-1) 77.

Table 77 – Emphasis flag definition

The FLc channel shows whether FLc (front left centre) channel data exist or not, as shown in [Table](#page-88-2) 78.

Table 78 – FLc channel definition

The FRc channel shows whether FRc (front right centre) channel data exist or not, as shown in [Table](#page-88-3) 79.

Table 79 – FRc channel definition

The LFE2 channel shows whether LFE2 (low frequency effects-2) channel data exist or not, as shown in [Table](#page-88-4) 80.

Table 80 – LFE2 channel definition

The BC channel shows whether BC (back centre) channel data exist or not, as shown in [Table](#page-89-0) 81.

The SiL channel shows whether SiL (side left) channel data exist or not, as shown in [Table](#page-89-1) 82.

Table 82 – SiL channel definition

Value	Description
	SiL (side left) channel data do not exist
	SiL (side left) channel data exist

The SiR channel shows whether SiR (side right) channel data exist or not, as shown in [Table](#page-89-2) 83.

Table 83 – SiR channel definition

The TpFL channel shows whether TpFL (top front left) channel data exist or not, as shown in [Table](#page-89-3) 84.

Table 84 – TpFL channel definition

The TpFR channel shows whether TpFR (top front right) channel data exist or not, as shown in [Table](#page-89-4) 85.

Table 85 – TpFR channel definition

12.5.9 Data transmitted at every data block of Group 3 for the fixed channels' structure

These ancillary data, as shown in [Figure](#page-90-0) 77 and [Table](#page-90-1) 86, are transmitted at every data block of Group 3 for the fixed channels' structure.

Figure 77 – Data transmitted at every data block of Group 3 for the fixed channels' structure

Table 86 – Data transmitted at every data block of Group 3 for the fixed channels' structure

The Emphasis flag shows whether de-emphasis is required for the sink device or not, as shown i[nTable](#page-90-2) 87.

Table 87 – Emphasis flag definition

The FLw channel shows whether FLw (front left wide) channel data exist or not, as shown in [Table](#page-90-3) 88.

Table 88 – FLw channel definition

The FRw channel shows whether FRw (front right wide) channel data exist or not, as shown in [Table](#page-90-4) 89.

Table 89 – FRw channel definition

The TpFC channel shows whether TpFC (top front centre) channel data exist or not, as shown in [Table](#page-91-0) 90.

Table 90 – TpFC channel definition

Value	Description
	TpFC (top front centre) channel data do not exist
	TpFC (top front centre) channel data exist

The TpC channel shows whether TpC (top centre) channel data exist or not, as shown in [Table](#page-91-1) 91.

The TpBL channel shows whether TpBL (top back left) channel data exist or not, as shown in [Table](#page-91-2) 92.

The TpBR channel shows whether TpBR (top back right) channel data exist or not, as shown in [Table](#page-91-3) 93.

Table 93 – TpBR channel definition

Value	Description
	TpBR (top back right) channel data do not exist
	TpBR (top back right) channel data exist

The TpSiL channel shows whether TpSiL (top side left) channel data exist or not, as shown in [Table](#page-91-4) 94.

The TpSiR channel shows whether TpSiR (top side right) channel data exist or not, as shown in [Table](#page-92-0) 95.

Table 95 – TpSiR channel definition

12.5.10 Data transmitted at every data block of Group 4 for the fixed channels' structure

These ancillary data, as shown in [Figure](#page-92-1) 78 and [Table](#page-92-2) 96, are transmitted at every data block of Group 4 for the fixed channels' structure.

Figure 78 – Data transmitted at every data block of Group 4 for the fixed channels' structure

Table 96 – Data transmitted at every data block of Group 4 for the fixed channels' structure

The Emphasis Flag shows whether de-emphasis is required for the sink device or not, as shown in [Table](#page-92-3) 97.

Table 97 – Emphasis flag definition

The TpBC channel shows whether TpBC (top back centre) channel data exist or not, as shown in [Table](#page-93-0) 98.

Table 98 – TpBC channel definition

The BtFC channel shows whether BtFC (bottom front centre) channel data exist or not, as shown in [Table](#page-93-1) 99.

The BtFL channel shows whether BtFL (bottom front left) channel data exist or not, as shown in [Table](#page-93-2) 100.

Table 100 – BtFL channel definition

The BtFR channel shows whether BtFR (bottom front right) channel data exist or not, as shown in [Table](#page-93-3) 101.

Table 101 – BtFR channel definition

The LSd channel shows whether LSd (left surround direct) channel data exist or not, as shown in [Table](#page-93-4) 102.

Table 102 – LSd channel definition

The RSd channel shows whether RSd (right surround direct) channel data exist or not, as shown in [Table](#page-93-5) 103.

Table 103 – RSd channel definition

The TpLS channel shows whether TpLS (top left surround) channel data exist or not, as shown in [Table](#page-94-0) 104.

Table 104 – TpLS channel definition

The TpRS channel shows whether TpRS (top right surround) channel data exist or not, as shown in [Table](#page-94-1) 105.

Table 105 – TpRS channel definition

Value	Description
U_{α}	TpRS (top right surround) channel data do not exist
	TpRS (top right surround) channel data exist

12.5.11 Data transmitted at every data block for the variable channels' structure

These ancillary data, as shown in [Figure](#page-94-2) 79 and [Table](#page-94-3) 106, are transmitted at every data block for the variable channels' structure.

Figure 79 – Data transmitted at every data block for the variable channel's structure

Table 106 – Data transmitted at every data block for the variable channels' structure

Data	Bits	Description
1) Emphasis Flag		Emphasis on or off
2) FL channel		FL (front left) channel data exist or not exist
3) FR channel		FR (front right) channel data exist or not exist
4) LFE1 channel		LFE1 (low frequency effects-1) channel data exist or not exist
5) FC channel		FC (front centre) channel data exist or not exist
6) LS channel		LS (left surround) channel data exist or not exist
7) RS channel		RS (right surround) channel data exist or not exist
8) BL channel		BL (back left) channel data exist or not exist
9) BR channel		BR (back right) channel data exist or not exist
10) FL/FR ch identifier	2	FL (front left)/FR (front right) channel identifier defined
11) FC ch identifier		FC (front centre) channel identifier defined
12) Extension ch flag 1		Extension channel bit order 1 exists or does not exist
13) Extension ch flag 2		Extension channel bit order 2 exists or does not exist

The emphasis flag shows whether de-emphasis is required for the sink device or not, as shown in [Table](#page-95-0) 107.

Table 107 – Emphasis flag definition

The FL channel shows whether FL (front left) channel data exist or not, as shown in [Table](#page-95-1) 108.

Table 108 – FL channel definition

Value	Description	
	FL (front left) channel data do not exist	
	FL (front left) channel data exist	

The FR channel shows whether FR (front right) channel data exist or not, as shown in [Table](#page-95-2) 109.

Table 109 – FR channel definition

The LFE1 channel shows whether LFE1 (low frequency effects-1) channel data exist or not, as shown in [Table](#page-95-3) 110.

Table 110 – LFE1 channel definition

The FC channel shows whether FC (front centre) channel data exist or not, as shown in [Table](#page-95-4) 111.

Table 111 – FC channel definition

The LS channel shows whether LS (left surround) channel data exist or not, as shown in [Table](#page-96-0) 112.

Table 112 – LS channel definition

The RS channel shows whether RS (right surround) channel data exist or not, as shown in [Table](#page-96-1) 113.

The BL channel shows whether BL (back left) channel data exist or not, see [Table](#page-96-2) 114.

Table 114 – BL channel definition

Value	Description	
v.	BL (back left) channel data do not exist	
\sim	BL (back left) channel data exist	

The BR channel shows whether BR (back right) channel data exist or not, as shown in [Table](#page-96-3) 115.

Table 115 – BR channel definition

The FL/FR ch identifier shows whether FL (front left)/FR (front right) channel data is FL/FR signal (stereo) or M1 (mono) signal or Lo (left output)/Ro (right output) signal or Lt (left total)/Rt (right total) signal, see [Table](#page-96-4) 116.

Note that the sink device shall decrease the M1 (mono) signal to a −3 dB level if the sink device outputs M1 (mono) signal in the L (left)/R (right) channel.

Table 116 – FL/FR ch identifier definition

The FC ch identifier shows whether FC (front centre) channel data is the FC signal or the M1 (mono) signal, as shown in [Table](#page-97-0) 117.

Table 117 – FC ch identifier definition

Value	Description	
υ.	FC (front centre) channel data is the FC signal	
\sim	FC (front centre) channel data is the M1 (mono) signal	

The extension ch flag 1 shows whether ancillary data of data transmitted at the extension channel bit order 1 for the variable channels' structure exist or not, as shown in [Table](#page-97-1) 118.

Table 118 – Extension ch flag 1 definition

Value	Description
U_2	Ancillary data of data transmitted at extension channel bit order 1 for the variable channels' structure do not exist
۱,	Ancillary data of data transmitted at extension channel bit order 1 for the variable channels' structure exist

The extension ch flag 2 shows whether ancillary data of data transmitted at extension channel bit order 2 for the variable channels' structure exist or not, as shown in [Table](#page-97-2) 119.

Table 119 – Extension ch flag 2 definition

Value	Description
0_{Ω}	Ancillary data of data transmitted at extension channel bit order 2 for the variable channels' structure do not exist
1_{2}	Ancillary data of data transmitted at extension channel bit order 2 for the variable channels' structure exist

12.5.12 Data transmitted at extension channel bit order 1 for the variable channels' structure

These ancillary data, as shown in [Figure](#page-97-3) 80 and [Table](#page-98-0) 120, are transmitted at extension channel bit order 1 for the variable channels' structure.

These data shall be transmitted as second ancillary data at least once per less than 100 ms when these data exist.

The transmitter should output these ancillary data as soon as possible to clarify the channel assignment as soon as content of the stream has been changed or the stream output has started.

Figure 80 – Data transmitted at extension channel bit order 1 for the variable channels' structure

Table 120 – Data transmitted at extension channel bit order 1 for the variable channels' structure

The FLc channel shows whether FLc (front left centre) channel data exist or not, as shown in [Table](#page-98-1) 121.

Table 121 – FLc channel definition

The FRc channel shows whether FRc (front right centre) channel data exist or not, as shown in, as shown in [Table](#page-98-2) 122.

Table 122 – FRc channel definition

The LFE2 channel shows whether LFE2 (low frequency effects-2) channel data exist or not, as shown in [Table](#page-98-3) 123.

Table 123 – LFE2 channel definition

The BC channel shows whether BC (back centre) channel data exist or not, as shown in [Table](#page-99-0) 124.

Table 124 – BC channel definition

The SiL channel shows whether SiL (side left) channel data exist or not, as shown in [Table](#page-99-1) 125.

Table 125 – SiL channel definition

The SiR channel shows whether SiR (side right) channel data exist or not, as shown in [Table](#page-99-2) 126.

Table 126 – SiR channel definition

The TpFL channel shows whether TpFL (top front left) channel data exist or not, as shown in [Table](#page-99-3) 127.

Table 127 – TpFL channel definition

The TpFR channel shows whether TpFR (top front right) channel data exist or not [Table](#page-99-4) 128.

Table 128 – TpFR channel definition

The FLw channel shows whether FLw (front left wide) channel data exist or not, as shown in [Table](#page-100-0) 129.

The FRw channel shows whether FRw (front right wide) channel data exist or not, as shown in [Table](#page-100-1) 130.

The TpFC channel shows whether TpFC (top front centre) channel data exist or not, as shown in [Table](#page-100-2) 131.

Table 131 – TpFC channel definition

The TpC channel shows whether TpC (top centre) channel data exist or not, as shown in $Table 132.$ $Table 132.$

Table 132 – TpC channel definition

The TpBL channel shows whether TpBL (top back left) channel data exist or not, as shown in [Table](#page-100-4) 133.

Table 133 – TpBL channel definition

The TpBR channel shows whether TpBR (top back right) channel data exist or not, as shown in [Table](#page-100-5) 134.

Table 134 – TpBR channel definition

The TpSiL channel shows whether TpSiL (top side left) channel data exist or not, as shown in [Table](#page-101-0) 135.

Table 135 – TpSiL channel definition

The TpSiR channel shows whether TpSiR (top side right) channel data exist or not, as shown in [Table](#page-101-1) 136.

12.5.13 Data transmitted at extension channel bit order 2 for the variable channels' structure

These ancillary data, as shown in [Figure](#page-101-2) 81 and [Table](#page-101-3) 137, are transmitted at extension channel bit order 2 for the variable channels' structure.

These data shall be transmitted as second ancillary data at least once per less than 100 ms provided that these data exist.

The transmitter should output these ancillary data as soon as possible to clarify the channel assignment as soon as content of the stream has been changed or the stream output has started.

Figure 81 – Data transmitted at extension channel bit order 2 for the variable channels' structure

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The TpBC channel shows whether TpBC (top back centre) channel data exist or not , as shown in [Table](#page-102-0) 138.

Table 138 – TpBC channel definition

Value	Description	
υ.,	TpBC (top back centre) channel data do not exist	
	TpBC (top back centre) channel data exist	

The BtFC channel shows whether BtFC (bottom front centre) channel data exist or not, as shown in [Table](#page-102-1) 139.

The BtFL channel shows whether BtFL (bottom front left) channel data exist or not, as shown in [Table](#page-102-2) 140.

Table 140 – BtFL channel definition

The BtFR channel shows whether BtFR (bottom front right) channel data exist or not, as shown in [Table](#page-102-3) 141.

Table 141 – BtFR channel definition

Value	Description	
	BtFR (bottom front right) channel data do not exist	
	BtFR (bottom front right) channel data exist	

The LSd channel shows whether LSd (left surround direct) channel data exist or not, as shown in [Table](#page-102-4) 142.

Table 142 – LSd channel definition

The RSd channel shows whether RSd (right surround direct) channel data exist or not, as shown in [Table](#page-103-0) 143.

The TpLS channel shows whether TpLS (top left surround) channel data exist or not, as shown in [Table](#page-103-1) 144.

Table 144 – TpLS channel definition

Value	Description	
U.	TpLS (top left surround) channel data do not exist	
\sim	TpLS (top left surround) channel data exist	

The TpRS channel shows whether TpRS (top right surround) channel data exist or not, as shown in [Table](#page-103-2) 145.

12.5.14 Data for CCI

[Figure](#page-103-3) 82 and [Table](#page-103-4) 146 provide a definition of ancillary data for CCI.

SUB LABEL $CO₁₆$ is for CCI.

Figure 82 – Ancillary data for CCI

Table 146 – Data transmitted at every data block

Data	Bits
1) CP-bit	
2) Category code	8
6) CGMS-A	2
5) CGMS-A validity	

NOTE 1 Ancillary data for CCI have the same meaning as specified in IEC [60958-3](http://dx.doi.org/10.3403/02859696U).

NOTE 2 Each value is based on the following condition in IEC [60958-3](http://dx.doi.org/10.3403/02859696U):

- consumer use of channel status block;
- audio sample word represents linear PCM samples;
- channel status mode is Mode 0.

These data shall be transmitted at least once during a 100 ms period.

12.5.15 Example of MBLA stream for the fixed channels' structure

[Figure](#page-105-0) 83 shows one-channel examples.

a) Example 1

$D4_{16}$	01_{16}	Ancillary Data
$D4_{16}$	CO ₁₆	Ancillary Data
42_{16}		Set to 000000_{16}
42_{16}	Set to 000000_{16}	
42_{16}	Set to 000000_{16}	
42_{16}		Mono
42_{16}		Set to 000000_{16}
42_{16}		Set to 000000_{16}
42_{16}		Set to 000000_{16}

b) Example 2

c) Example 3

Figure 83 – Examples of MBLA stream for the fixed channels' structure of one channel

The channel structure shown in example 2 of [Figure](#page-105-0) 83 appears when only FL and FR are transmitted.

[Figure](#page-105-1) 84 shows two-channel examples.

a) Example 1

b) Example 2

Figure 84 – Examples of MBLA stream for the fixed channels' structure of two channels

[Figure](#page-106-0) 85 shows a three-channel example.

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Figure 85 – Example of MBLA stream for the fixed channels' structure of three channels (3/0)

[Figure](#page-106-1) 86 shows a four-channel example.

Figure 86 – Example of MBLA stream for the fixed channels' structure of four channels (2/2)

12.5.16 Example of MBLA stream for the variable channels' structure

[Figure](#page-107-0) 87 shows one-channel examples.

a) Example 1

b) Example 2

Figure 87 – Examples of MBLA stream for the variable channels' structure of one channel

[Figure](#page-107-1) 88 shows a two-channel example.

Figure 88 – Example of MBLA stream for the variable channels' structure two channels

[Figure](#page-107-2) 89 shows a three-channel example.

Figure 89 – Example of MBLA stream for the variable channels' structure of three channels (3/0)

[Figure](#page-108-0) 90 shows a four-channel example.

Figure 90 – Example of MBLA stream for the variable channels' structure of four channels (2/2)

[Figure](#page-108-0) 91 shows a seven-channel example.

$\rm D4_{16}$	01_{16}	Ancillary Data
$D4_{16}$	CO ₁₆	Ancillary Data
42_{16}	Front Left	
42_{16}	Front Right	
42_{16}	Front Centre	
42_{16}	Left Surround	
42_{16}	Right Surround	
42_{16}	Front Left wide	
42_{16}	Front Right wide	
CF_{16}	CF_{16}	

Figure 91 – Example of MBLA stream for the fixed channels' structure of seven channels

Annex A

(informative)

Synchronization issues

A.1 General

The following synchronization issues have been identified:

- a) rate matching between the transmitter and receiver;
- b) adjusting the presentation time at a receiver;
- c) adjusting the location at a transmitter.

The rate matching between the transmitter and receiver can be done by one of two methods:

- 1) clock-based rate control;
- 2) command-based rate control (see [10.4\)](#page-42-0).

Clock-based rate control may use sampling clock delivery in an isochronous stream or another clock delivery system such as a dedicated clock.

The presentation time adjustment of the application sequence at a receiver can be done since the time stamp of a CIP is defined in such a way that it reflects the time when the corresponding audio sample goes out of a buffer for depacketization. If an application requires precise adjustment of the presentation time, the application should take into account the extra delay caused by signal processing or A/D and D/A conversion.

A.2 Delivery of sampling clock of arbitrary frequency

This clause focuses on rate matching in terms of sampling clock delivery, which is very familiar to audio engineers. It only applies to real time transfer, which occurs when the sample transmission frequency is used to define the sampling frequency.

Since a CIP without a source packet header (SPH) has only one time stamp in the SYT field, the maximum synchronization clock frequency shall be limited to the isochronous cycle of 8 kHz.

Assume that a transmitter carries an audio stream with sampling frequency *STF* and that *STF* > 8 kHz.

The transmitter derives a synchronization clock with frequency F_{sync} according to Equation (A.1):

$$
F_{\text{sync}} = STF / \text{SYT_INTERVAL} < 8\,000\tag{A.1}
$$

where

The transmitter quantizes the timing of the synchronization clock, for instance, the rising edge of the clock, by referring to its own CYCLE_TIME. It transmits the sum of the timing and TRANSFER_DELAY by using the SYT field of the CIP. The resolution of the time stamp is 1/(24,576 MHz), or approximately 40 ns, and CYCLE_TIME may have 40 ns of jitter due to this quantization. If the timing information is not available for a CIP, the SYT shall indicate the no information code.

A receiver can reproduce the synchronization clock in terms of the pulse generated when the SYT equals its own CYCLE TIME.

The sampling clock can be reproduced by multiplying the synchronization clock by the SYT_INTERVAL, which shall be determined before receiving begins.

This sampling clock delivery does not require synchronization of the sampling clock and the isochronous cycle.

The reproduced synchronization clock will have jitter. This jitter can degrade audio quality unless adequate jitter attenuation is used.

The local CYCLE_TIME registers at the transmitter and the receiver nodes will have jitter from various sources. This CYCLE TIME register jitter has a minimum peak-peak amplitude equivalent to the approximately 40 ns resolution of CYCLE TIME. If one of the nodes is the cycle master this jitter only applies to CYCLE_TIME at the other node. If neither of the nodes is the cycle master then it will apply to CYCLE TIME at both the transmit and received nodes. There is also a source of CYCLE TIME jitter from the quantization of the correction for variable delay to the cycle start packets from the cycle master.

The jitter added to the synchronization clock by delivery in this manner is the sum of the CYCLE TIME jitter and the jitter due to the quantization of the time stamp.

Annex B

(informative)

Catching up in non-blocking transmission method

Equation (3) in [7.4.1](#page-22-0) stipulates that in normal operation, each transmitter shall construct a packet containing between 0 and SYT_INTERVAL events. [Table](#page-39-0) 20 specifies SYT_Interval for each sample transmission frequency such that: provide

Event arrival time SYT_INTERVAL-11 – Event arrival time [0] > Min_period (B.1)

and

125 μ s <= Min Period

where

Event arrival time[M] is the time (measured in μ s) of the arrival at the transmitter of event at index M. The event with index $= 0$ is the event which has Presentation Time = SYT. *Min Period* **bigget is the time (measured in us) of SYT INTERVAL events.**

The Min Period ensures that at most only a single SYT will be required for each packet.

Using a normal non-blocking transmission method, fewer SYT_INTERVAL events will be transmitted in each packet. In the event of lost opportunities to transmit a packet (such as a cycle start packet drop after a bus reset) a transmitter can catch up by transmitting up to SYT_INTERVAL events in one or more of the subsequent packets. Events which are late according to Equation (4) are not transmitted.

Equation (9) in [9.2](#page-38-0) can be used to determine the required isochronous bandwidth, but, in a normal non-blocking operation, not all of this bandwidth is used. Extra bandwidth is available for catching up. However, this extra bandwidth may not be sufficient to ensure that some events will not be late.

A method is provided below to allow a transmitter to add one extra event to each catch-up packet as long as the total number of events is not greater than the SYT_INTERVAL.

In Equation (9) in [9.2,](#page-38-0) the term (int(max(Fs)/8 000) + 1) can be changed to (int(max(Fs)/8 000) + 2). This increases the allocated bandwidth in such a way that one additional event can be sent per packet. While this bandwidth will be unused during normal operation, it will provide the extra bandwidth needed to catch up without violating the allocated bandwidth.

It is important to consider that in the case of lost isochronous cycles, more than one transmitter may be trying to catch up at the same time. Sufficient bandwidth should be allocated to allow for catch-up.

Annex C

(informative)

Transport characteristics

C.1 Sampling-clock jitter characteristics

Sampling-clock jitter can degrade the accuracy of conversion processes in sampling devices. This part of the annex describes the jitter mechanisms in the exchange of sample timing information and derives worst-case jitter levels to be used for stressing sampling devices when making performance measurements.

This issue applies to systems that require a sample clock to be transferred across the bus to a sampling device. For example, it does not apply for devices that use flow control with a single sampling device acting as destination and synchronization master, or where the destination device is a non-sampling device such as a recorder.

C.2 Sample clock transfer jitter mechanisms using A/M protocol

C.2.1 General

The A/M protocol and the serial bus use asynchronous clocks to define and exchange timing and synchronization information. The changing phase relationships and limited timing resolution of these clocks, and, in some circumstances, the changing phase relationship to an external sample clock, produce a variable error which introduces jitter into an embedded synchronization clock.

There are other sources of jitter including oscillator phase noise, variable gate delays and cable inter-symbol interference. These are normally small in comparison with the mechanisms considered here.

C.2.2 CYCLE_TIME register jitter

C.2.2.1 General

Embedded synchronization clock information is referenced to the CYCLE_TIME register value at the synchronization clock source. Jitter on this register value at the synchronization clock source and synchronization clock destination nodes contributes to embedded synchronization clock jitter.

C.2.2.2 Cycle start packet CYCLE_TIME resolution

The cycle start packet issued from the cycle master is used to align the CYCLE_TIME registers of any isochronous-capable nodes on a serial bus. It is transmitted at or after the cycle counter on the cycle master node and is incremented. It carries the value of the cycle master node CYCLE_TIME register at the time the cycle start is initiated.

Asynchronous activity on the bus at the time when the cycle starts an event causes a delay in transmitting the cycle start packet. At the other isochronous nodes, the CYCLE_TIME register is loaded with the value carried on the cycle start packet. That compensates for the cycle start delay but only up to the resolution of that register. This resolution is 1/24,576 MHz (which is approximated in this annex as 41 ns).

The cycle start packet carries a value from the CYCLE_TIME register. If the transmission of the packet is timed so that it always occurs at a fixed time after the moment that the CYCLE TIME register is updated to that value, then the cycle start delays will be corrected without a significant error. This means that asynchronous activity at the time when the cycle starts an event will not generate jitter.

However, some IEEE 1394 compliant implementations might introduce a variable delay between the time the CYCLE_TIME register is updated and the cycle start packet transmission of that value. This will depend on the implementation, but this delay may be limited to less than the 41 ns CYCLE TIME resolution, or it could possibly be even greater than this.

C.2.2.3 Variable transport delay to cycle start packets

As a cycle start packet is passed through intermediate nodes on the bus it is delayed by a variable amount of repeater data delay.

The normal mechanism for the variation in this delay is the re-timing of the packet by the local clock at each node. The repeater data delay varies as the relative timing of the incoming transitions and the local clock changes. This change is a result of the frequency difference between the local clock and the clock on the previous node the packet has passed through. Jitter produced in this way is in the form of a ramping variation with a step correction in the opposite direction. The frequency of this sawtooth is related to the frequency difference between the two node clocks.

IEEE 1394 does not define explicit limits for repeater delay jitter. The draft supplement, P1394a, specifies a PHY register field jitter that can indicate values from 1/49,152 MHz (which is approximated in this annex as 20 ns) to 7/49,152 MHz (approximately 163 ns).

IEEE 1394 PHY devices that resynchronize received data with a 49,152 MHz clock will have a repeater data delay jitter of approximately 20 ns peak-peak or 6 ns r.m.s.

The jitter due to a variable repeater delay jitter is cumulative. The total variable transport delay is the sum of the delay at each node. The total r.m.s. jitter to the cycle start packet transport delay is the root sum of squares (RSS) of the r.m.s. jitter at each intermediate repeater node.

C.2.2.4 Quantization of CYCLE_TIME register correction

The CYCLE TIME registers at each isochronous node increment at a rate defined by the exact rate of the 24,576 MHz clock in the local node. These registers are time aligned with similar registers in other nodes by being loaded with the value carried in the cycle start packet transmitted by the cycle master. As the CYCLE_TIME register incrementing clock has a slightly different frequency at each node, there will be a gradually changing error between the updating of that register at the cycle master and the other nodes.

When there is a difference between the value on an incoming cycle start packet and the value in the local CYCLE TIME register then a correction is made.

This correction is quantized to the CYCLE_TIME register resolution of 1/24,576 MHz. The contribution of this mechanism to the CYCLE_TIME register jitter is normally a gradually increasing delay or advance with corrective step in the opposite direction. This jitter has an amplitude equivalent to the CYCLE_TIME resolution of 41 ns peak-to-peak and 12 ns r.m.s.

C.2.3 Time-stamp quantization jitter

The time stamp (SYT) carrying the sampling timing information has a resolution of 1/24,576 MHz. The effect of quantization to this resolution is to add jitter to the embedded sample clock. This jitter has an amplitude equivalent to the SYT resolution of 41 ns peak-topeak and 12 ns r.m.s. It will have frequency components related to the beat frequency between the time stamp rate (Fs/SYT INTERVAL) and the 24,576 MHz clock incrementing the CYCLE_TIME register.

C.3 Embedded sample-clock jitter

C.3.1 Embedded sample-clock jitter spectrum

The error in the values and timing of the embedded synchronization clock can be considered as a time-varying signal. This can be examined in the frequency domain through spectrum analysis. This jitter spectrum will relate to the jitter spectrum in the sample clock transfer mechanism and the jitter transfer function.

There are discrete frequency components corresponding to the fundamental and harmonic frequencies associated with each of the applicable jitter sources described in the previous subclause. These frequencies depend on the frequency differences between the local PHY clocks on the nodes.

Any jitter source that produces a jitter signal similar to a sawtooth will have discrete jitter frequency components at the sawtooth frequency and multiples of that rate. Where the multiple is at a frequency above half the frequency that the timing information is updated, then that component will be aliased to below that rate and the signal will no longer appear as a sawtooth.

C.3.2 Embedded sample-clock jitter amplitude

C.3.2.1 General

The total amount of embedded sample-clock jitter is dependent on the following:

- the number of nodes between the cycle master and sample-clock source;
- the number of nodes between the cycle master and sample-clock destination;
- the implementation of each node;
- whether or not the sample-clock source is synchronized to the bus.

C.3.2.2 Example one: simple two-node bus

As an example, examine the simplest two-node system as shown in [Figure](#page-114-0) C.1. This two-node bus has the cycle master as the sample clock source node (node 0), and the sample clock is locked to the sample clock source node PHY clock at a multiple of the cycle time rate. Asynchronous activity is low enough to ensure that the cycle start packet is never delayed.

Figure C.1 – Two-node bus

The jitter analysis of this example is as follows.

- There will not be any jitter due to cycle start packet CYCLE_TIME resolution as the cycle start packet is not delayed due to asynchronous activity.
- There is no variable transport delay to cycle start packets as there are no intermediate nodes on the bus.
- Quantization of CYCLE_TIME register correction in the sample clock destination node will be a source of jitter in this example. This will be in the form of one sawtooth at a frequency determined by the offset between the cycle start rate and the sample clock destination PHY clock. This will have an amplitude of approximately 12 ns r.m.s. (41 ns peak to peak).
- As the sample clock is frequency-locked to the cycle master PHY clock, there is no timestamp quantization jitter.

Therefore, for the simple two-node system in this example, the recovered embedded sample clock will have just one systematic jitter source. This will have a jitter amplitude of approximately 12 ns r.m.s. (41 ns peak to peak) in the form of a sawtooth at a rate determined by the frequency offset between the two PHY node clocks.

C.3.2.3 Example two: three-node bus

In this example, there are three nodes which are: the cycle master node, the sample clock source node and the sample clock destination node, see [Figure](#page-115-0) C.2.

Figure C.2 – Three-node bus

The following analysis also assumes that the sample clock is not synchronous to any of the bus clocks.

- If the cycle start packet is sometimes delayed, there may be some jitter caused when the cycle start packet CYCLE_TIME value does not exactly correspond to the delay of the packet transmission. This will have a peak amplitude that is dependent on the implementation of the cycle master cycle start transmission mechanism. (The amplitude of this mechanism is not included in the analysis.)
- In the path from cycle master (node 0) to node 1 there are no intermediate nodes. In the path from cycle master (node 0) to node 2, there is one intermediate node that will have a variable transport delay to cycle start packets. This will contribute to the jitter in the CYCLE TIME value at that node. This jitter will be in the form of a sawtooth related to the beating of node 0 and node 1 cycle clocks. The amplitude of this jitter mechanism depends on the implementation of the repeater function in this node. This analysis assumes that this repeater includes resynchronization with a 49,152 MHz clock. This will contribute jitter of approximately 6 ns r.m.s. (20 ns peak to peak).
- Quantization of CYCLE_TIME register correction in nodes 1 and 2 will be a source of jitter. In each of these nodes, this will be in the form of a sawtooth at a frequency determined by the offset between the cycle start rate and the node PHY clock. These two sources of jitter will each have an amplitude of approximately 12 ns r.m.s. (41 ns peak to peak).
- At node 1, the sample clock timing is encoded into the SYT with the resolution of the CYCLE TIME register. The sample clock is asynchronous to the update of the CYCLE_TIME register. The error due to the variation in relative phase of the clocks is a sawtooth with a frequency determined by the difference between the node 1 cycle clock frequency and the time stamp rate. This source of jitter will have an amplitude of approximately 12 ns r.m.s. (41 ns peak to peak).

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The list above illustrates that this system has four sources of periodic jitter (excluding the source of jitter related to asynchronous activity): three of 12 ns r.m.s. and one of 6 ns r.m.s. The sum total of the periodic jitter (excluding the component due to asynchronous activity) will be 21 ns r.m.s. (This would also have a peak-to-peak value of 132 ns. This value represents the infrequent coincidence of the peaks of all the contributing jitter components and would be an infrequent occurrence.)

C.3.2.4 Example: thirty-five-node system

This example illustrates a large bus configuration with 23 hops between the cycle master (node 0) and each sample clock source (node 23) and sample clock destination (node 34), see [Figure](#page-116-0) C.3. (According to IEEE 1394A, this configuration represents a maximum within the constraints of a maximum PHY delay of 144 ns and maximum cable length of 4,5 m.)

Figure C.3 – Thirty-five-node bus

The following analysis also makes similar assumptions for the 3-node example with respect to the sample clock.

- If there is asynchronous activity on the bus, then the jitter mechanism due to the cycle start packet delay is the same as for the three-node example. This is not included in the analysis.
- In the paths from the cycle master (node 0) to both the sample clock source (node 23) and sample clock destination (node 34) there are 22 intermediate nodes. Each of these will impose a variable transport delay on to cycle start packets in the same manner as the 3-node example. The peak jitter will scale in proportion to the number of hops (22) and the r.m.s. jitter will scale with the square root of that number, 4,7. If each repeater applies resynchronization with a local 49,152 MHz clock, then they will add a total of 28 ns r.m.s. of jitter to the arrival time of the cycle start packet at the sample clock source (node 23) and at the sample clock destination (node 34).
- As with the 3-node example, quantization of CYCLE_TIME register correction at the sample clock source and destination will be a source of jitter of amplitude 12 ns r.m.s. each.
- As with the 3-node example, the time-stamp quantization jitter will add 12 ns r.m.s.

This illustrates how this system has three sources of periodic sawtooth jitter at 12 ns r.m.s. and two summed periodic components at 28 ns r.m.s. each. The sum total of the periodic jitter is 44 ns r.m.s.

This result does not represent a 'worst case'. The variable transport delay jitter at each intermediate node could be significantly greater than 20 ns while remaining compliant with IEEE 1394. The potential variable error in the CYCLE_TIME value in the cycle start packet (when the cycle start has been delayed by asynchronous activity) has also not been included.

C.4 Jitter attenuation

Jitter attenuation occurs with the filtering function of the sample-clock recovery device. This will have a low pass jitter attenuation characteristic. Sample-clock jitter causes modulation of the sampled signal. These modulation products may become audible. For high-quality applications, it is recommended that the jitter attenuation characteristic of the sample-clock recovery system satisfies the template shown in [Figure](#page-117-0) C.4.

Figure C.4 – Sample-clock recovery jitter attenuation template

To satisfy this template, the jitter attenuation plotted against jitter frequency shall fall below the shaded subclauses of the graph. The attenuation shall exceed 60 dB at jitter frequencies above 200 Hz and up to half the recovered sample clock frequency. Below 200 Hz the gain shall not exceed 3 dB.

The jitter attenuation for received jitter at frequencies, *f*^r above half the SYT_MATCH clock rate, f_s is determined by the response to the images of the received jitter that may be present in the sampling clock. These will be present at image frequencies of:

$$
f_{\mathsf{i}} = N \times f_{\mathsf{S}} \pm f_{\mathsf{r}}
$$

where *N* is an integer.

C.5 Jitter measurement

Jitter meters approximate the long-term average frequency and phase of a signal that they are measuring. This will result in a high-pass characteristic. As the sample clocks derived using the A/M protocol have a strong low-frequency jitter component, the low-frequency corner frequency of the jitter meter is important.

It is recommended that jitter measurements use the characteristics defined by the jitter measurement filter characteristic of [Figure](#page-118-0) C.5.

Figure C.5 – Sample clock jitter measurement filter characteristic

This is a minimum-phase high pass filter with a -3 dB frequency of 700 Hz, a first order rolloff to 70 Hz and with a pass-band gain of unity.

NOTE This is compatible with the intrinsic jitter measurement filter characteristic used in [IEC 60958-3](http://dx.doi.org/10.3403/02859696U) and [IEC 60958-4](http://dx.doi.org/10.3403/02509544U).

Bibliography

- [1] [ISO/IEC 13213:1994](http://dx.doi.org/10.3403/00539181), *Information technology Microprocessor system Control and Status Register (CSR) Architecture for microcomputer buses*
- [2] MIDI 1.0, *The Complete MIDI 1.0 Detailed Specification*
- [3] MMA/AMEI RP-027, *Specification for MIDI Media Adaptation Layer for IEEE-1394* Version 1.0. Available at [<http://www.midi.org/>](http://www.midi.org/) and [http://www.amei.or.jp/>](http://www.amei.or.jp/)
- [4] *ASID Specification, ASID Specification (Audio Software Information Delivery)* Version 1.0, IFPI, RIAA and RIAJ. Available at [<http://www.riaa.org>](http://www.riaa.org/) and [<http://www.riaj.org>](http://www.riaj.org/)
- [5] AES3-2009, *AES standard for digital audio engineering Serial transmission format for two-channel linearly represented digital audio dat*a. Available at <http://www.aes.org>
- [6] AES58-2008, *AES standard for digital audio Audio applications of networks Application of [IEC 61883-6](http://dx.doi.org/10.3403/02755535U) 32-bit generic data*. Available at <http://www.aes.org>
- [7] 1394 Trade Association document 1999014, *Enhancement to Audio and Music Data Transmission Protocol 1.0.* Available at [<http://www.1394ta.org>](http://www.1394ta.org/)
- [8] 1394 Trade Association document 1999015, *AV/C Command Set for Rate Control of Isochronous Data Flow 1.0.* Available at [<http://www.1394ta.org>](http://www.1394ta.org/)
- [9] 1394 Trade Association document 1999024, *SMPTE Time Code and Sample Count Transmission Protocol Version 1.0.* Available at [<http://www.1394ta.org>](http://www.1394ta.org/)
- [10] 1394 *Trade Association document 1999026, AV/C Digital Interface Command Set General Specification Version 4.0. Available at [<http://www.1394ta.org>](http://www.1394ta.org/)*
- [11] 1394 Trade Association document 2009013, Audio and Music Data Transmission Protocol 2.2.l Available at [<http://www.1394ta.org>](http://www.1394ta.org/)
- [12] Super Audio CD System Description Version 1.2
- [13] DVD Specification for Read-Only Disc Part 4, Audio Specifications Version 1.0 March 1999
- [14] DVD Specification for Read-Only Disc Part 4, Audio Specifications Version-up Information (from 1.1 to 1.2) May 2000
- [15] System Description Blu-ray Disc Read-Only Format Part 3: Audio Visual Basic Specifications Version 2.5 June 2011
- [16] SMPTE 0428-3-2006, D-Cinema Distribution Master Audio Channel Mapping and Channel Labelling
- [17] [IEC 60958-4](http://dx.doi.org/10.3403/02509544U), *Digital audio interface Part 4: Professional applications*
- [18] [IEC 61883-6:2002](http://dx.doi.org/10.3403/02755535), *Consumer audio/video equipment Digital interface Part 6: Audio and music data transmission protocol*[3](#page-119-0)

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³ This reference is given for the sake of backwards compatibility.

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